

## LECTURE NOTES 7

### EM WAVE PROPAGATION IN CONDUCTORS

Inside a conductor, free charges can move/migrate around in response to  $EM$  fields contained therein, as we saw for the case of the longitudinal  $\vec{E}$ -field inside a current-carrying wire that had a static potential difference  $\Delta V$  across its ends. Even in the static case of electric charge residing on the surface of a conductor, we saw that  $\vec{E}_{inside}(\vec{r}) = 0$ , but recall that this actually means (as we showed last semester) that the NET electric field inside the conductor is zero, *i.e.*  $\vec{E}_{inside}^{NET}(\vec{r}) = 0$ .

*n.b.* here, we assume {for simplicity's sake} that the conductor is linear/homogeneous/isotropic – *i.e.* no crystalline structure/no anisotropies/no inhomogenities/voids/defects...

From Ohm's Law, we know that the free current density  $\vec{J}_{free}(\vec{r}, t)$  is proportional to the (ambient) electric field inside the conductor:  $\vec{J}_{free}(\vec{r}, t) = \sigma_C \vec{E}(\vec{r}, t)$  where:

$\sigma_C = \text{conductivity}$  of the metal conductor ( $\text{Siemens}/m = \text{Ohm}^{-1}/m$ ) and  $\sigma_C = 1/\rho_C$   
 $\rho_C = \text{resistivity}$  of the metal conductor ( $\text{Ohm}\cdot m$ ).

Thus inside such a conductor, we can assume that the linear/homogeneous/isotropic conducting medium has electric permittivity  $\epsilon$  and magnetic permeability  $\mu$ . Maxwell's equations inside such a conductor {with  $\vec{J}_{free}(\vec{r}, t) \neq 0$ } are thus:

1) $\vec{\nabla} \cdot \vec{E}(\vec{r}, t) = \rho_{free}(\vec{r}, t) / \epsilon$	2) $\vec{\nabla} \cdot \vec{B}(\vec{r}, t) = 0$	Using Ohm's Law: $\vec{J}_{free}(\vec{r}, t) = \sigma_C \vec{E}(\vec{r}, t)$
3) $\vec{\nabla} \times \vec{E}(\vec{r}, t) = -\frac{\partial \vec{B}(\vec{r}, t)}{\partial t}$	4) $\vec{\nabla} \times \vec{B}(\vec{r}, t) = \mu \vec{J}_{free}(\vec{r}, t) + \mu \epsilon \frac{\partial \vec{E}(\vec{r}, t)}{\partial t} = \mu \sigma_C \vec{E}(\vec{r}, t) + \mu \epsilon \frac{\partial \vec{E}(\vec{r}, t)}{\partial t}$	

Electric charge is (always) conserved, thus the continuity equation inside the conductor is:

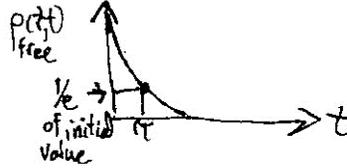
$\vec{\nabla} \cdot \vec{J}_{free}(\vec{r}, t) = -\frac{\partial \rho_{free}(\vec{r}, t)}{\partial t}$	but: $\vec{J}_{free}(\vec{r}, t) = \sigma_C \vec{E}(\vec{r}, t)$	
$\therefore \sigma_C (\vec{\nabla} \cdot \vec{E}(\vec{r}, t)) = -\frac{\partial \rho_{free}(\vec{r}, t)}{\partial t}$	but: $\vec{\nabla} \cdot \vec{E}(\vec{r}, t) = \frac{1}{\epsilon} \rho_{free}(\vec{r}, t)$	
thus: $\frac{\sigma_C \rho_{free}(\vec{r}, t)}{\epsilon} = -\frac{\partial \rho_{free}(\vec{r}, t)}{\partial t}$	or: $\frac{\partial \rho_{free}(\vec{r}, t)}{\partial t} + \left(\frac{\sigma_C}{\epsilon}\right) \rho_{free}(\vec{r}, t) = 0$	1 <sup>st</sup> order linear, homogeneous differential equation

The general solution of this differential equation for the free charge density is of the form:

$$\rho_{free}(\vec{r}, t) = \rho_{free}(\vec{r}, t=0) e^{-\sigma_C t / \epsilon} = \rho_{free}(\vec{r}, t=0) e^{-t / \tau_{relax}} \quad \text{i.e. a damped exponential !!!}$$

Characteristic damping time:  $\tau_{relax} \equiv \epsilon / \sigma_C = \text{charge relaxation time}$  {aka time constant}.

Thus, the continuity equation  $\vec{\nabla} \cdot \vec{J}_{free}(\vec{r}, t) = -\partial \rho_{free}(\vec{r}, t) / \partial t$  inside a conductor tells us that any free charge density  $\rho_{free}(\vec{r}, t=0)$  initially present at time  $t=0$  is **exponentially** damped / dissipated in a characteristic time  $\tau_{relax} \equiv \varepsilon / \sigma_C = \text{charge relaxation time \{aka time constant\}}$ , such that at when:  $t = \tau_{relax} \equiv \varepsilon / \sigma_C : \rho_{free}(\vec{r}, t = \tau_{relax}) = \rho_{free}(\vec{r}, t=0) e^{-1} = 0.369 \cdot \rho_{free}(\vec{r}, t=0)$



### Calculation of the Charge Relaxation Time for Pure Copper:

$$\rho_{Cu} = 1 / \sigma_{Cu} = 1.68 \times 10^{-8} \Omega \cdot m \Rightarrow \sigma_{Cu} = 1 / \rho_{Cu} = 5.95 \times 10^7 \text{ Siemens/m}$$

If we assume  $\varepsilon_{Cu} \approx 3\varepsilon_o = 3 \times 8.85 \times 10^{-8} \text{ F/m}$  for copper metal, then:

$$\tau_{Cu}^{relax} = \varepsilon_{Cu} / \sigma_{Cu} = \rho_{Cu} \varepsilon_{Cu} = 4.5 \times 10^{-19} \text{ sec} \quad !!!$$

However, the characteristic (aka mean) **collision time** of free electrons in pure copper is  $\tau_{Cu}^{coll} \approx \lambda_{Cu}^{coll} / v_{thermal}^{Cu}$  where  $\lambda_{Cu}^{coll} \approx 3.9 \times 10^{-8} \text{ m}$  = mean free path (between successive collisions) in pure copper, and  $v_{thermal}^{Cu} \approx \sqrt{3k_B T / m_e} \approx 12 \times 10^5 \text{ m/sec}$  and thus we obtain:  $\tau_{coll}^{Cu} \approx 3.2 \times 10^{-13} \text{ sec}$ .

Hence we see that the calculated charge relaxation time in pure copper,  $\tau_{Cu}^{relax} \approx 4.5 \times 10^{-19} \text{ sec}$  is  $\ll$  than the calculated collision time in pure copper,  $\tau_{coll}^{Cu} \approx 3.2 \times 10^{-13} \text{ sec}$ .

Furthermore, the **experimentally measured** charge relaxation time in pure copper is  $\tau_{Cu}^{relax}(\text{exp't}) \approx 4.0 \times 10^{-14} \text{ sec}$ , which is  $\approx 5$  orders of magnitude **larger** than the **calculated** charge relaxation time  $\tau_{Cu}^{relax} \approx 4.5 \times 10^{-19} \text{ sec}$ . The problem here is that {the **macroscopic**} Ohm's Law is simply out of its range of validity on such short time scales! Two **additional** facts here are that **both**  $\varepsilon$  and  $\sigma_C$  are **frequency-dependent** quantities {i.e.  $\varepsilon = \varepsilon(\omega)$  and  $\sigma_C = \sigma_C(\omega)$ }, which becomes **increasingly** important at the higher frequencies ( $f = 2\pi/\omega \sim 1/\tau_{relax}$ ) associated with short time-scale, transient-type phenomena!

So in reality, if we are willing to wait a short time (e.g.  $\Delta t \sim 1 \text{ ps} = 10^{-12} \text{ sec}$ ) then, any initial free charge density  $\rho_{free}(\vec{r}, t=0)$  accumulated inside a **good** conductor at  $t=0$  will have dissipated away/damped out, and from that time onwards,  $\rho_{free}(\vec{r}, t) = 0$  **can** be safely assumed.

Note: For a **poor** conductor ( $\sigma_C \rightarrow 0$ ), then:  $\tau_{relax} \equiv \varepsilon / \sigma_C \rightarrow \infty$  !!! Please keep this in mind...

After **many** charge relaxation time constants, e.g.  $20\tau_{relax} \leq \Delta t \approx 1 \text{ ps} = 10^{-12} \text{ sec}$ , Maxwell's **steady-state** equations for a **good** conductor become {with  $\rho_{free}(\vec{r}, t \geq \Delta t) = 0$  from then onwards}:

1) $\vec{\nabla} \cdot \vec{E}(\vec{r}, t) = 0$	2) $\vec{\nabla} \cdot \vec{B}(\vec{r}, t) = 0$	Maxwell's equations for a <b>charge-equilibrated</b> conductor
3) $\vec{\nabla} \times \vec{E}(\vec{r}, t) = -\frac{\partial \vec{B}(\vec{r}, t)}{\partial t}$		
4) $\vec{\nabla} \times \vec{B}(\vec{r}, t) = \mu\sigma_c \vec{E}(\vec{r}, t) + \mu\epsilon \frac{\partial \vec{E}(\vec{r}, t)}{\partial t} = \mu \left( \sigma_c \vec{E}(\vec{r}, t) + \epsilon \frac{\partial \vec{E}(\vec{r}, t)}{\partial t} \right)$		

These equations are different from the previous derivation(s) of monochromatic plane *EM* waves propagating in free space/vacuum and/or in linear/homogeneous/isotropic non-conducting materials {n.b. only equation 4) has changed}, hence we re-derive {**steady-state**} wave equations for  $\vec{E}$  &  $\vec{B}$  from scratch. As before, we apply  $\vec{\nabla} \times ( )$  to equations 3) and 4):

$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = -\frac{\partial}{\partial t} (\vec{\nabla} \times \vec{B})$	$\vec{\nabla} \times (\vec{\nabla} \times \vec{B}) = \mu \left( \sigma_c (\vec{\nabla} \times \vec{E}) \right) + \epsilon \frac{\partial}{\partial t} (\vec{\nabla} \times \vec{E})$
$= \vec{\nabla} (\cancel{\vec{\nabla} \cdot \vec{E}}) - \nabla^2 \vec{E} = -\frac{\partial}{\partial t} \left( \mu\sigma_c \vec{E} + \mu\epsilon \frac{\partial \vec{E}}{\partial t} \right)$	$= \vec{\nabla} (\cancel{\vec{\nabla} \cdot \vec{B}}) - \nabla^2 \vec{B} = -\mu\sigma_c \frac{\partial \vec{B}}{\partial t} - \mu\epsilon \frac{\partial^2 \vec{B}}{\partial t^2}$
$= \nabla^2 \vec{E} = \mu\epsilon \frac{\partial^2 \vec{E}}{\partial t^2} + \mu\sigma_c \frac{\partial \vec{E}}{\partial t}$	$= \nabla^2 \vec{B} = \mu\epsilon \frac{\partial^2 \vec{B}}{\partial t^2} + \mu\sigma_c \frac{\partial \vec{B}}{\partial t}$
<u>Again:</u> $\nabla^2 \vec{E}(\vec{r}, t) = \mu\epsilon \frac{\partial^2 \vec{E}(\vec{r}, t)}{\partial t^2} + \mu\sigma_c \frac{\partial \vec{E}(\vec{r}, t)}{\partial t}$	<u>and:</u> $\nabla^2 \vec{B}(\vec{r}, t) = \mu\epsilon \frac{\partial^2 \vec{B}(\vec{r}, t)}{\partial t^2} + \mu\sigma_c \frac{\partial \vec{B}(\vec{r}, t)}{\partial t}$

Note that the {**steady-state**} 3-D wave equations for  $\vec{E}$  and  $\vec{B}$  in a conductor have an additional term that has a single time derivative – which is analogous e.g. to a velocity-dependent **damping term** associated with the motion of a 1-D mechanical harmonic oscillator.

The general solution(s) to the above {**steady-state**} wave equations are usually in the form of an oscillatory function  $\times$  a damping term (i.e. a decaying exponential) – in the direction of the propagation of the *EM* wave, complex plane-wave type solutions for  $\vec{E}$  and  $\vec{B}$  associated with the above wave equation(s) are of the general form:

$\vec{\tilde{E}}(\vec{r}, t) = \vec{\tilde{E}}_o e^{i(\vec{k}z - \omega t)}$	and:	$\vec{\tilde{B}}(\vec{r}, t) = \vec{\tilde{B}}_o e^{i(\vec{k}z - \omega t)} = \frac{1}{\tilde{v}} \hat{k} \times \vec{\tilde{E}}(\vec{r}, t) = \left( \frac{\tilde{k}}{\omega} \right) \hat{k} \times \vec{\tilde{E}}(\vec{r}, t)$
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n.b. with {frequency-dependent} **complex** wave number:  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega)$

where:  $k(\omega) = \Re\{\tilde{k}(\omega)\}$  and  $\kappa(\omega) = \Im\{\tilde{k}(\omega)\}$  and corresponding **complex** wave vector

$\tilde{\hat{k}}(\omega) = \tilde{k}(\omega) \hat{k} = \tilde{k}(\omega) \hat{z}$  (for *EM* wave propagating in the  $\hat{k} = +\hat{z}$  direction, **here**).

Physically,  $k(\omega) = \Re\{\tilde{k}(\omega)\}$  is associated with wave **propagation**, and  $\kappa(\omega) = \Im\{\tilde{k}(\omega)\}$  is associated with wave **attenuation** (i.e. **dissipation**).

We plug  $\tilde{\vec{E}}(\vec{r}, t) = \tilde{E}_o e^{i(\tilde{k}z - \omega t)}$  and  $\tilde{\vec{B}}(\vec{r}, t) = \tilde{B}_o e^{i(\tilde{k}z - \omega t)}$  into their respective wave equations above, and obtain from each wave equation the same/identical **characteristic equation** – {aka a **dispersion relation**} between complex  $\tilde{k}(\omega)$  and  $\omega$  {please work this out yourselves!}:

$$\boxed{\tilde{k}^2(\omega) = \mu\epsilon\omega^2 + i\mu\sigma_c\omega}$$

Thus, since  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega)$ , then:

$$\boxed{\tilde{k}^2(\omega) = (k(\omega) + i\kappa(\omega))^2 = k^2(\omega) - \kappa^2(\omega) + 2ik(\omega)\kappa(\omega) = \mu\epsilon\omega^2 + i\mu\sigma_c\omega}$$

If we {temporarily} suppress the  $\omega$ -dependence of complex  $\tilde{k}(\omega)$ , this relation becomes:

$$\boxed{\tilde{k}^2 = (k + i\kappa)^2 = k^2 - \kappa^2 + 2ik\kappa = \mu\epsilon\omega^2 + i\mu\sigma_c\omega}$$

We can re-write this expression as:  $\boxed{[(k^2 - \kappa^2) - \mu\epsilon\omega^2] + i[2k\kappa - \mu\sigma_c\omega] = 0}$ , which **must** be true for **any/all** values of {any of} the parameters involved. The only in-general way that this relation can hold is if **both**  $[(k^2 - \kappa^2) - \mu\epsilon\omega^2] = 0$  **and**,  $[2k\kappa - \mu\sigma_c\omega] = 0$ . Then:

$$\boxed{k^2 - \kappa^2 = \mu\epsilon\omega^2} \quad \text{and:} \quad \boxed{2k\kappa = \mu\sigma_c\omega}$$

Thus, we have **two** separate/independent equations:  $k^2 - \kappa^2 = \mu\epsilon\omega^2$  and:  $2k\kappa = \mu\sigma_c\omega$ . We have **two** unknowns:  $k$  and  $\kappa$ . Hence, we solve these equations **simultaneously** to determine  $k$  and  $\kappa$ !

From the **latter** relation, we see that:  $\boxed{\kappa = \frac{1}{2}\mu\sigma_c\omega/k}$ . Plug **this** result into the **other** relation:

$$\boxed{k^2 - \kappa^2 = k^2 - \left(\frac{1}{2}\mu\sigma_c\omega/k\right)^2 = k^2 - \frac{1}{k^2}\left(\frac{1}{2}\mu\sigma_c\omega\right)^2 = \mu\epsilon\omega^2}$$

Then multiply by  $k^2$  and rearrange the terms to obtain the following relation:

$$\boxed{k^4 - (\mu\epsilon\omega^2)k^2 - \left(\frac{1}{2}\mu\sigma_c\omega\right)^2 = 0}$$

This may **look** like a scary equation to try to solve (*i.e.* a **quartic** equation - *eeekkk!*), but it's actually just a **quadratic** equation! {So, it's really just a **leprechaun**, masquerading as a **unicorn!**}

**Define:**  $x \equiv k^2$ ,  $a \equiv 1$ ,  $b \equiv -(\mu\epsilon\omega^2)$  and  $c \equiv -\left(\frac{1}{2}\mu\sigma_c\omega\right)^2$ , this equation then becomes

“the usual” quadratic equation, of the form:  $ax^2 + bx + c = 0$ , with solution(s)/root(s):

$$\boxed{x = \frac{-b \mp \sqrt{b^2 - 4ac}}{2a}} \quad \text{or:} \quad \boxed{k^2 = \frac{1}{2} \left[ +(\mu\epsilon\omega^2) \mp \sqrt{(\mu\epsilon\omega^2)^2 + 4\left(\frac{1}{2}\mu\sigma_c\omega\right)^2} \right]}$$

This relation can be re-written as:

$$k^2 = \frac{1}{2}(\mu\epsilon\omega^2) \left[ 1 \mp \sqrt{1 + \frac{(\mu^2 \sigma_c^2 \omega^2)}{4(\mu^2 \epsilon^2 \omega^4)}} \right] = \frac{1}{2}(\mu\epsilon\omega^2) \left[ 1 \mp \sqrt{1 + \frac{(\sigma_c^2)}{(\epsilon^2 \omega^2)}} \right] = \frac{1}{2}(\mu\epsilon\omega^2) \left[ 1 \mp \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} \right]$$

On **physical** grounds ( $k^2 > 0$ ), we **must** select the + sign, hence:

$$k^2 = \frac{1}{2}(\mu\epsilon\omega^2) \left[ 1 + \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} \right] \text{ and thus: } k = \sqrt{k^2} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ 1 + \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} \right]^{1/2} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{1/2}$$

Having thus solved for  $k$  (or equivalently,  $k^2$ ), we can use **either** of our original **two** relations to solve for  $\kappa$ , e.g.  $k^2 - \kappa^2 = \mu\epsilon\omega^2$ , thus:

$$\kappa^2 = k^2 - \mu\epsilon\omega^2 = \frac{1}{2}(\mu\epsilon\omega^2) \left[ 1 + \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} \right] - \mu\epsilon\omega^2 = \frac{1}{2}(\mu\epsilon\omega^2) \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} - 1 \right]$$

Hence {finally}, we obtain:

$$k(\omega) = \Re\{ \tilde{k}(\omega) \} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{1/2} \text{ and: } \kappa(\omega) = \Im\{ \tilde{k}(\omega) \} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} - 1 \right]^{1/2}$$

The above two relations **clearly** show the frequency dependence of **both** the **real** and **imaginary** components of the complex wavenumber  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega)$ . This physically means that EM wave propagation in a conductor is **dispersive** (i.e. EM wave propagation is **frequency dependent**).

Note also that the **imaginary** part of  $\tilde{k}(\omega)$ ,  $\kappa(\omega) = \Im\{ \tilde{k}(\omega) \}$  results in an **exponential attenuation/damping** of the monochromatic plane EM wave with increasing  $z$ :

$$\begin{aligned} \tilde{E}(\vec{r}, t) &= \tilde{E}_o e^{i(\tilde{k}z - \omega t)} = \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \quad \text{where: } \tilde{k}(\omega) = k(\omega) + i\kappa(\omega) \\ \text{and: } \tilde{B}(\vec{r}, t) &= \tilde{B}_o e^{i(\tilde{k}z - \omega t)} = \tilde{B}_o e^{-\kappa z} e^{i(kz - \omega t)} = \frac{\tilde{k}}{\omega} \hat{k} \times \tilde{E}(z, t) = \frac{\tilde{k}}{\omega} \hat{k} \times \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \end{aligned}$$

The **characteristic distance**  $z$  over which  $\tilde{E}$  and  $\tilde{B}$  are attenuated/reduced to  $1/e = e^{-1} = 0.368$  of their initial values (at  $z = 0$ ) is known as the **skin depth**,  $\delta_{sc}(\omega) \equiv 1/\kappa(\omega)$  (SI units: meters).

$$\text{i.e. } \delta_{sc}(\omega) = \frac{1}{\kappa(\omega)} = \frac{1}{\omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} - 1 \right]^{1/2}} \Rightarrow \begin{aligned} \tilde{E}(z = \delta_{sc}, t) &= \tilde{E}_o e^{-1} e^{i(kz - \omega t)} \\ \tilde{B}(z = \delta_{sc}, t) &= \tilde{B}_o e^{-1} e^{i(kz - \omega t)} \end{aligned}$$

The **real** part of  $\tilde{k}(\omega)$ , i.e.  $k(\omega) = \Re\{\tilde{k}(\omega)\}$  determines the **spatial** wavelength  $\lambda(\omega)$ , the **phase** speed  $v_\phi(\omega)$  and also the **group** speed  $v_g(\omega)$  of the monochromatic *EM* plane wave in the conductor:

$$\lambda(\omega) = \frac{2\pi}{k(\omega)} = \frac{2\pi}{\Re\{\tilde{k}(\omega)\}}$$

$$v_\phi(\omega) \equiv \frac{\omega}{k(\omega)} = \frac{\omega}{\Re\{\tilde{k}(\omega)\}}$$

$v_\phi(\omega)$  = propagation speed of a **point** on waveform that has **constant phase**  $\Phi$ .

**Phase**  $\Phi \equiv (kz - \omega t) = \text{constant}$ .  
A constant phase **point** on the waveform moves:  $z(t) = \Phi/k + v_\phi t$ .

$$v_g(\omega) \equiv \frac{1}{dk(\omega)/d\omega} = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1}$$

$v_g(\omega)$  = propagation speed of **energy** / **information**.

We will discuss **phase** speed  $v_\phi(\omega)$  and the **group** speed  $v_g(\omega)$  more – later...

The above plane wave solutions satisfy the above *EM* wave equations(s) for **any** choice of  $\tilde{\vec{E}}_o$ . As we have also seen before, it can similarly be shown here that Maxwell's equations 1) and 2) ( $\vec{\nabla} \cdot \vec{E} = 0$  and  $\vec{\nabla} \cdot \vec{B} = 0$ ) rule out the presence of any **longitudinal**  $z$ -components for  $\vec{E}$  and  $\vec{B}$ .  
 $\Rightarrow$  For *EM* waves propagating in a conductor,  $\vec{E}$  and  $\vec{B}$  are {still} **purely transverse**!

If we consider e.g. a **linearly polarized** monochromatic plane *EM* wave propagating in the  $\hat{k} = +\hat{z}$  -direction in a conducting medium, e.g.  $\tilde{\vec{E}}(\vec{r}, t) = \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{x}$ , then:

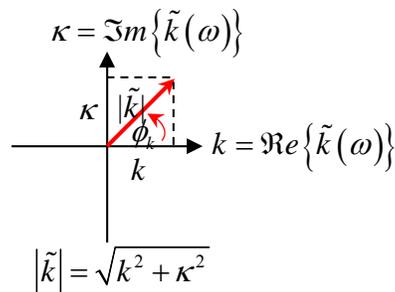
$$\tilde{\vec{B}}(\vec{r}, t) = \left( \frac{\tilde{k}}{\omega} \right) \hat{k} \times \tilde{\vec{E}}(\vec{r}, t) = \left( \frac{\tilde{k}}{\omega} \right) \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y} = \left( \frac{k + i\kappa}{\omega} \right) \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y}$$

$$\Rightarrow \tilde{\vec{E}}(\vec{r}, t) \perp \tilde{\vec{B}}(\vec{r}, t) \perp (\hat{k} = +\hat{z}) \quad (\hat{k} = +\hat{z} = \text{propagation direction of } EM \text{ wave, } \underline{\text{here}})$$

The complex wavenumber:  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega) = |\tilde{k}(\omega)| e^{i\phi_k(\omega)}$

where:  $|\tilde{k}(\omega)| = \sqrt{\tilde{k}(\omega) \tilde{k}^*(\omega)} = \sqrt{k^2(\omega) + \kappa^2(\omega)}$  and:  $\phi_k(\omega) = \tan^{-1}(\kappa(\omega)/k(\omega))$

In the complex  $\tilde{k}$ -plane:



Then we see that:  $\tilde{\vec{E}}(\vec{r}, t) = \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{x}$  has:  $\tilde{E}_o = E_o e^{i\delta_E}$   $\tilde{k} = |\tilde{k}| e^{i\phi_k}$

and that:  $\tilde{\vec{B}}(\vec{r}, t) = \tilde{B}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y} = \frac{\tilde{k}}{\omega} \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y}$  has:  $\tilde{B}_o = B_o e^{i\delta_B} = \frac{\tilde{k}}{\omega} \tilde{E}_o = \frac{|\tilde{k}| e^{i\phi_k}}{\omega} E_o e^{i\delta_E}$

Thus, we see that:  $B_o e^{i\delta_B} = \frac{|\tilde{k}| e^{i\phi_k}}{\omega} E_o e^{i\delta_E} = \frac{|\tilde{k}|}{\omega} E_o e^{i(\delta_E + \phi_k)} = \frac{\sqrt{k^2 + \kappa^2}}{\omega} E_o e^{i(\delta_E + \phi_k)}$

i.e., inside a conductor,  $\vec{E}$  and  $\vec{B}$  are **no longer in phase with each other!!!**

Phases of  $\vec{E}$  and  $\vec{B}$ :  $\delta_B = \delta_E + \phi_k$

With phase **difference**:  $\Delta\phi_{B-E} \equiv \delta_B - \delta_E = \phi_k \leftarrow$  magnetic field **lags** behind electric field!!!

We also see that:  $\frac{B_o}{E_o} = \frac{|\tilde{k}|}{\omega} = \left[ \epsilon\mu \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} \right]^{1/2} \neq \frac{1}{c}$

The real/physical  $\vec{E}$  and  $\vec{B}$  fields associated with linearly polarized monochromatic plane EM waves propagating in a conducting medium are **exponentially** damped:

$\vec{E}(\vec{r}, t) = \Re e \left\{ \tilde{\vec{E}}(\vec{r}, t) \right\} = E_o e^{-\kappa z} \cos(kz - \omega t + \delta_E) \hat{x}$   $\delta_B = \delta_E + \phi_k$

$\vec{B}(\vec{r}, t) = \Re e \left\{ \tilde{\vec{B}}(\vec{r}, t) \right\} = B_o e^{-\kappa z} \cos(kz - \omega t + \delta_B) \hat{y} = B_o e^{-\kappa z} \cos(kz - \omega t + \{\delta_E + \phi_k\}) \hat{y}$

$\frac{B_o}{E_o} = \frac{|\tilde{k}(\omega)|}{\omega} = \left[ \epsilon\mu \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} \right]^{1/2}$  where:  $|\tilde{k}(\omega)| = \sqrt{k^2(\omega) + \kappa^2(\omega)} = \omega \left[ \epsilon\mu \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} \right]^{1/2}$

$\delta_B = \delta_E + \phi_k$ ,  $\phi_k(\omega) \equiv \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right)$  and:  $\tilde{k}(\omega) = (k(\omega) + i\kappa(\omega)) \hat{z}$ ,  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega)$

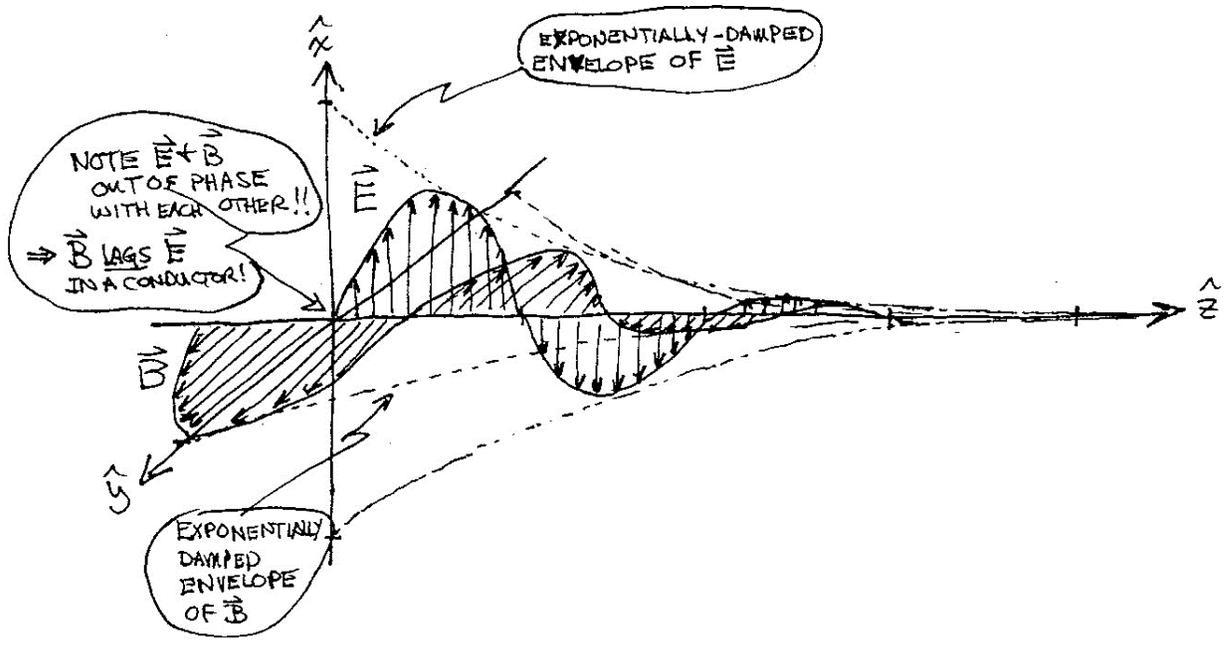
Definition of the **skin depth**  $\delta_{sc}(\omega)$  in a conductor:

$\delta_{sc}(\omega) \equiv \frac{1}{\kappa(\omega)} = \frac{1}{\omega \sqrt{\frac{\epsilon\mu}{2} \left[ \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} - 1 \right]^{1/2}}} =$  Distance  $z$  over which the  $\vec{E}$  and  $\vec{B}$  fields fall to  $1/e = e^{-1} = 0.368$  of their initial values.

The **instantaneous** power **per unit volume** in the conductor {ultimately dissipated as **heat!**} is:

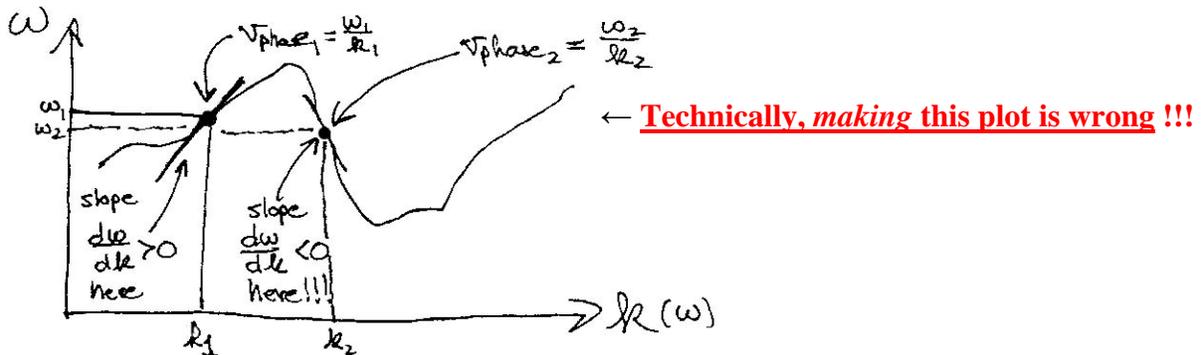
$p(\vec{r}, t) = \vec{J}(\vec{r}, t) \cdot \vec{E}(\vec{r}, t) = \sigma_c \vec{E}(\vec{r}, t) \cdot \vec{E}(\vec{r}, t) = \sigma_c E^2(\vec{r}, t) = E_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E) \text{ (Watts/m}^3\text{)}$

The **time-averaged** power **per unit volume** in the conductor is:  $\langle p(\vec{r}, t) \rangle_t = \frac{1}{2} E_o^2 e^{-2\kappa z} \equiv p_o e^{-\alpha z}$



**Phase Speed vs. Group Speed of a Wave:**

The **phase** speed = **numerical value** of  $v_\phi(\omega) \equiv \omega/k(\omega)$  at a **point** on the  $\omega$  vs.  $k(\omega)$  curve,  
 The **group** speed  $v_g(\omega) \equiv "d\omega/dk(\omega)"$  = the **local slope** at a **point** on the  $\omega$  vs.  $k(\omega)$  curve:



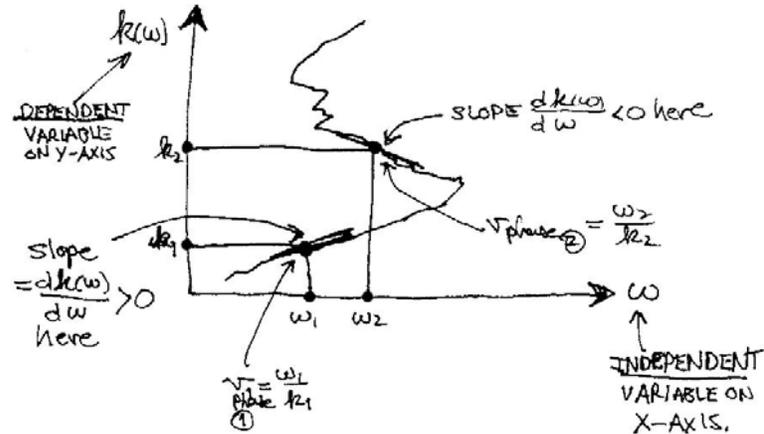
**Why is this plot technically wrong ???** **Because:**  $k(\omega) = fcn(\omega)$

i.e.  $\omega$  is the **independent** variable {**always** plotted on the **axis of abscissas** (i.e. the **x-axis**)}  
 $k(\omega)$  is the **dependent** variable {**always** plotted on the **axis of ordinates** (i.e. the **y-axis**)}

Thus, the **technically correct** way **is** to plot  $k(\omega)$  vs.  $\omega$ : {because  $k$  depends on  $\omega$ , **not** vice-versa!}

Then:  $v_g(\omega) \equiv 1 / \left( \frac{dk(\omega)}{d\omega} \right) = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1}$  i.e.  $v_g(\omega) \equiv 1 / \text{slope of } \{k(\omega) \text{ vs. } \omega\} \text{ graph}$

See below:



Another way to think about this issue is to remember that the angular frequency  $\omega$  and wavenumber  $k(\omega)$  are **Fourier transforms** of time  $t$  and position  $z(t)$ , respectively. In the **space-time domain**, clearly the space position  $z(t)$  is the **dependent** variable, time  $t$  is the **independent** variable. The Fourier transform of the **dependent** variable  $z(t)$  is the **dependent** variable  $k(\omega)$ , the Fourier transform of the **independent** variable  $t$  is the **independent** variable  $\omega$ .

Thus **here**, for the physics associated with propagation of *EM* plane waves in a conductor, with frequency-dependent **real**-component wavenumber  $k(\omega)$ :

$$k(\omega) = \Re\{\tilde{k}(\omega)\} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{1/2}$$

The **phase** speed:

$$v_\phi(\omega) \equiv \frac{\omega}{k(\omega)} = \frac{1}{\sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{1/2}}$$

The **group** speed:

$$v_g(\omega) \equiv \frac{1}{dk(\omega)/d\omega} = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1} = \left[ \frac{d}{d\omega} \left\{ \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{1/2} \right\} \right]^{-1}$$

So let's work out what the **group speed**  $v_g(\omega)$  is for an *EM* plane wave propagating in a conductor.

Using the chain rule of differentiation:

$$\begin{aligned}
 \frac{dk(\omega)}{d\omega} &= \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} + \omega \sqrt{\frac{\epsilon\mu}{2}} \frac{d}{d\omega} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} \\
 &= \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} + \omega \sqrt{\frac{\epsilon\mu}{2}} \cdot \frac{\frac{1}{2} \cdot \frac{1}{\omega^2} \cdot \cancel{2} \cdot \left(\frac{\sigma_c}{\epsilon}\right)^2 \left(-\frac{1}{\omega^3}\right)}{\left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}} \\
 &= \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} - \frac{1}{2} \sqrt{\frac{\epsilon\mu}{2}} \cdot \frac{\left(\frac{\sigma_c}{\epsilon\omega}\right)^2}{\left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}} \\
 &= \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}} \cdot \left\{ 1 - \frac{1}{2} \frac{\left(\frac{\sigma_c}{\epsilon\omega}\right)^2}{\left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right] \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}} \right\}
 \end{aligned}$$

The **group speed** of an *EM* plane wave propagating in a conductor is: (eek!!!)

$$v_g(\omega) = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1} = \frac{1}{\sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}}} \cdot \frac{1}{\left\{ 1 - \frac{1}{2} \frac{\left(\frac{\sigma_c}{\epsilon\omega}\right)^2}{\left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right] \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}} \right\}}$$

The relation between **phase speed** vs. **group speed** of an *EM* plane wave propagating in a conductor is:

$$v_\phi(\omega) \equiv \frac{\omega}{k(\omega)} = \frac{1}{\sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right]^{\frac{1}{2}}} \quad v_g(\omega) = v_\phi(\omega) \cdot \frac{1}{\left\{ 1 - \frac{1}{2} \frac{\left(\frac{\sigma_c}{\epsilon\omega}\right)^2}{\left[ \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2} + 1 \right] \sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}} \right\}}$$

### EM Wave Complex Impedance in a Conductor:

The complex vector impedance associated with an *EM* wave propagating in a conductor is:

$$\tilde{Z}(\vec{r}, t; \omega) \equiv \tilde{\vec{E}}(\vec{r}, t; \omega) \times \tilde{\vec{H}}^{-1}(\vec{r}, t; \omega) = \frac{\tilde{\vec{E}}(\vec{r}, t; \omega) \times \tilde{\vec{H}}^*(\vec{r}, t; \omega)}{|\tilde{\vec{H}}(\vec{r}, t; \omega)|^2} = \mu \frac{\tilde{\vec{E}}(\vec{r}, t; \omega) \times \tilde{\vec{B}}^*(\vec{r}, t; \omega)}{|\tilde{\vec{B}}(\vec{r}, t; \omega)|^2}$$

If the electric and magnetic fields associated with the *EM* wave propagating in the conductor are:

$$\tilde{\vec{E}}(\vec{r}, t) = \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{x} \quad \text{and} \quad \tilde{\vec{B}}(\vec{r}, t) = \left( \frac{\tilde{k}}{\omega} \right) \hat{k} \times \tilde{\vec{E}}(\vec{r}, t) = \left( \frac{\tilde{k}}{\omega} \right) \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y}$$

Then:

$$\begin{aligned} \tilde{Z}(\vec{r}, t; \omega) &= \mu \frac{\tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{x} \times \left( \frac{\tilde{k}^*(\omega)}{\omega} \right) \tilde{E}_o^* e^{-\kappa z} e^{-i(kz - \omega t)} \hat{y}}{\left| \left( \frac{\tilde{k}(\omega)}{\omega} \right) \tilde{E}_o e^{-\kappa z} e^{i(kz - \omega t)} \hat{y} \right|^2} \\ &= \mu \omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \frac{|\tilde{E}_o|^2 e^{-2\kappa z}}{|\tilde{E}_o|^2 e^{-2\kappa z}} \hat{z} = \mu \omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z} \quad (\text{Ohms}) \end{aligned}$$

Note that {again}:  $\tilde{Z}(\vec{r}, t; \omega) = \mu \omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z}$  has **no** explicit time dependence

$$\tilde{Z}(\vec{r}, \omega) = \mu \omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu \omega \left( \frac{k(\omega) - i\kappa(\omega)}{k^2(\omega) + \kappa^2(\omega)} \right) \hat{z} \quad (\text{Ohms})$$

Complex impedance is manifestly a complex frequency-domain quantity

Note that since:  $\tilde{k}^*(\omega) = k(\omega) - i\kappa(\omega) = |\tilde{k}^*(\omega)| e^{-i\varphi_k(\omega)} = |\tilde{k}(\omega)| e^{-i\varphi_k(\omega)}$

and:  $\varphi_k(\omega) = \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right) = \delta_B(\omega) - \delta_E(\omega)$

We can also equivalently write this expression as:

$$\begin{aligned} \tilde{Z}(\vec{r}, \omega) &= \mu \omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu \omega \left( \frac{|\tilde{k}(\omega)| e^{-i\varphi_k(\omega)}}{|\tilde{k}(\omega)|^2} \right) \hat{z} \\ &= \mu \left( \frac{\omega}{|\tilde{k}(\omega)|} \right) e^{-i(\delta_B(\omega) - \delta_E(\omega))} \hat{z} = \mu \left( \frac{\omega}{|\tilde{k}(\omega)|} \right) e^{+i(\delta_E(\omega) - \delta_B(\omega))} \hat{z} \end{aligned}$$

**EM Wave Propagation in a Conductor – Special/Limiting Cases:**

 a) **Good conductors:**  $\sigma_c \gg \varepsilon\omega$  Conductivity of a **good** conductor:  $\sigma_c \rightarrow \infty$  (i.e.  $\rho_c = 1/\sigma_c \rightarrow 0$ ).

 Since:  $\tilde{k} = k + ik$  and:  $\sigma_c \gg \varepsilon\omega$ , i.e.  $\left(\frac{\sigma_c}{\varepsilon\omega} \gg 1\right)$  for a **good** conductor. Then:

$$k \equiv \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\varepsilon\omega}\right)^2} + 1 \right]^{1/2}} \approx \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\varepsilon\omega}\right)^2} \right]^{1/2}} \approx \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \frac{\sigma_c}{\varepsilon\omega} \right]^{1/2}} = \omega \sqrt{\frac{\cancel{\varepsilon} \mu \sigma_c}{2 \cancel{\varepsilon} \omega}} = \sqrt{\frac{\omega \mu \sigma_c}{2}}$$

and:

$$\kappa \equiv \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\varepsilon\omega}\right)^2} - 1 \right]^{1/2}} \approx \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \sqrt{1 + \left(\frac{\sigma_c}{\varepsilon\omega}\right)^2} \right]^{1/2}} \approx \omega \sqrt{\frac{\varepsilon\mu}{2} \left[ \frac{\sigma_c}{\varepsilon\omega} \right]^{1/2}} = \omega \sqrt{\frac{\cancel{\varepsilon} \mu \sigma_c}{2 \cancel{\varepsilon} \omega}} = \sqrt{\frac{\omega \mu \sigma_c}{2}}$$

 $\therefore$  In a **good** conductor  $\left(\frac{\sigma_c}{\varepsilon\omega} \gg 1\right)$ :

$$k(\omega) = \kappa(\omega) = \sqrt{\frac{\omega \mu \sigma_c}{2}} \quad \text{and skin depth:} \quad \delta_{sc}(\omega) \equiv \frac{1}{\kappa(\omega)} \approx \sqrt{\frac{2}{\omega \mu \sigma_c}}$$

**FORMULAS FOR EM WAVE PROPAGATION IN A GOOD CONDUCTOR**

$$\left(\frac{\sigma_c}{\varepsilon\omega} \gg 1\right)$$

$$k(\omega) = \kappa(\omega) = \sqrt{\frac{\omega \mu \sigma_c}{2}} \quad \text{and:} \quad \delta_{sc}(\omega) = \text{skin depth} \equiv \frac{1}{\kappa(\omega)} \approx \sqrt{\frac{2}{\omega \mu \sigma_c}}$$

$$\text{Wavenumber, } k(\omega) \equiv \frac{2\pi}{\lambda(\omega)} \Rightarrow \lambda(\omega) = \frac{2\pi}{k(\omega)} \approx \frac{2\pi}{\kappa(\omega)} = 2\pi \delta_{sc}(\omega) = 2\pi \sqrt{\frac{2}{\omega \mu \sigma_c}}$$

 n.b. in a **perfect** conductor:  $\sigma_c = \infty$ 

$$\Rightarrow k(\omega) = \kappa(\omega) = \sqrt{\frac{\omega \mu \sigma_c}{2}} = \infty$$

$$\Rightarrow \lambda(\omega) = \frac{2\pi}{k(\omega)} = 0$$

$$\delta_{sc}(\omega) = \frac{1}{\kappa(\omega)} \approx \sqrt{\frac{2}{\omega \mu \sigma_c}} = 0$$

$$\phi_k(\omega) \equiv (\delta_B - \delta_E) = \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right) \approx \tan^{-1}(1)$$

$$\text{But: } \tan^{-1}(1) = 45^\circ = \frac{\pi}{4}$$

$$\Rightarrow \phi = \delta_B - \delta_E = 45^\circ = \frac{\pi}{4}$$

 $\Rightarrow \vec{B}$  **lags**  $\vec{E}$  by  $\approx 45^\circ$  in a **good** conductor.

 n.b. In a **perfect** conductor:  $\sigma_c = \infty$ ,  $\phi \equiv 45^\circ = \frac{\pi}{4}$ 
**In a typical good conductor (e.g. gold/silver/copper/...):**  $\left(\frac{\sigma_c}{\varepsilon\omega} \gg 1\right)$

For **optical** frequencies/visible light region:  $\omega \approx 10^{16}$  radians/sec. A **good** conductor typically has  $\sigma_c \approx 10^7$  Siemens/m and  $\epsilon \approx 3\epsilon_0$ , and at optical frequencies:  $(\sigma_c/\epsilon\omega) \approx 37.7 \gg 1$  is satisfied.

If the conductor is  $\cong$  **non-magnetic** (e.g. copper, aluminum, gold, silver, platinum... etc.)  
 $\Rightarrow \mu \approx \mu_0 = 4\pi \times 10^{-7}$  Henrys/m.

Then:  $k(\omega) \approx \kappa(\omega) = \sqrt{\frac{\omega\mu\sigma_c}{2}} \approx \sqrt{\frac{\omega\mu_0\sigma_c}{2}} = \left[ \frac{10^{16} \times 4\pi \times 10^{-7} \times 10^7}{2} \right]^{1/2} \approx 2.51 \times 10^8 \text{ radians/m}$

And:  $\lambda(\omega) = 2\pi/k(\omega)$  = wavelength in good conductor  $\approx 2.51 \times 10^{-8} \text{ m} = 25.1 \text{ nm}$

cf w/ vacuum wavelength:  $\lambda_o = \frac{2\pi}{k_o} = \frac{2\pi c}{\omega} = \frac{c}{f} = \frac{2\pi \times 3 \times 10^8}{10^{16}} = 1.885 \times 10^{-7} \text{ m} = 188.5 \text{ nm}$

$\Rightarrow \lambda(\omega) \approx 25.1 \text{ nm} \left( \text{good conductor} \right) \ll \lambda_o = 188.5 \text{ nm} \left( \text{vacuum wavelength} \right)$

Vacuum/conductor  $\lambda$ -ratio:  $\left( \frac{\lambda_o}{\lambda(\omega)} \right) = \frac{188.5 \text{ nm}}{25.1 \text{ nm}} \approx 7.52$  at **optical** frequencies,  $\omega \approx 10^{16}$  rad/sec.

Skin depth:  $\delta_{sc}(\omega) = \frac{1}{\kappa(\omega)} \approx \frac{\lambda(\omega)}{2\pi} \approx 4.0 \times 10^{-9} \text{ m} = 4.0 \text{ nm} \quad !!!$

$\Rightarrow$  This explains why metals are **opaque** at optical frequencies,  $\omega \approx 10^{16}$  radians/sec  
 {and **also** e.g. explains why/how silvered sunglasses work!}

Compare these results for **EM** waves propagating in conductors at **optical** frequencies to those for **EM** waves propagating in conductors, but instead at **very low** frequencies – e.g. the AC line frequency,  $f_{AC} = 60 \text{ Hz} \Rightarrow \omega_{AC} = 2\pi f_{AC} = 120\pi$  rad/sec, where the criterion for a **good** conductor,  $(\sigma_c/\epsilon\omega) \approx 10^{15} \gg 1$  is certainly well-satisfied:

$$\text{At } f = 60 \text{ Hz: } \left\{ \begin{array}{l} k_{AC} \approx \kappa_{AC} \approx \sqrt{\frac{\omega\mu\sigma_c}{2}} = \left[ \frac{120\pi \times 4\pi \times 10^{-7} \times 10^7}{2} \right] = 48.7 \text{ radians/m} \\ \lambda_{AC} = \frac{2\pi}{k} = 0.129 \text{ m} = 12.9 \text{ cm} \\ \lambda_{o_{AC}} = 5 \times 10^6 \text{ m}!! \\ \frac{\lambda_{o_{AC}}}{\lambda_{AC}} = \frac{5 \times 10^6 \text{ m}}{0.129 \text{ m}} \approx 3.87 \times 10^7 !! \\ 60 \text{ Hz AC skin depth: } \delta_{sc}^{AC} = \frac{\lambda_{AC}}{2\pi} \approx 2.05 \times 10^{-2} \text{ m} = 2.05 \text{ cm}!! \end{array} \right.$$

$\Rightarrow$  Need **at least**  $3-4 \times \delta_{sc} \approx$  several  $\rightarrow 10 \text{ cm}$  to screen out unwanted 60 Hz AC signals !!!

### Phase speed vs. group speed in a Good conductor:

Given that:  $k(\omega) \approx \kappa(\omega) \approx \sqrt{\frac{\omega\mu\sigma_c}{2}}$  in a **good** conductor, where:  $\left(\frac{\sigma_c}{\varepsilon\omega}\right) \gg 1$

The **phase** speed and **group** speed in a **good** conductor are respectively:

$$v_\phi(\omega) \equiv \frac{\omega}{k(\omega)} \approx \frac{\omega}{\sqrt{\frac{\omega\mu\sigma_c}{2}}} \quad \text{and:} \quad v_g(\omega) = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1} = \frac{1}{\sqrt{\frac{\omega\mu\sigma_c}{2}}} \cdot \frac{1}{\left\{1 - \frac{1}{2}\right\}} = \frac{2}{\sqrt{\frac{\omega\mu\sigma_c}{2}}} = 2v_\phi(\omega) !!!$$

### Complex impedance in a Good conductor:

$$\tilde{Z}(\vec{r}, \omega) = \mu\omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu\omega \left( \frac{|\tilde{k}(\omega)| e^{-i\phi_k(\omega)}}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu \left( \frac{\omega}{|\tilde{k}(\omega)|} \right) e^{-i\phi_k(\omega)} \hat{z} \quad (\text{Ohms})$$

Since:  $k(\omega) \approx \kappa(\omega) \approx \sqrt{\frac{\mu\omega\sigma_c}{2}}$  Then:  $|\tilde{k}(\omega)| = \sqrt{k^2(\omega) + \kappa^2(\omega)} = \sqrt{2}k(\omega) \approx \sqrt{\mu\omega\sigma_c}$

And:  $\phi_k(\omega) = \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right) \approx \tan^{-1}(1) = 45^\circ = \delta_B(\omega) - \delta_E(\omega) \Rightarrow \vec{B} \text{ lags } \vec{E} \text{ by } \approx 45^\circ \text{ in a } \underline{\text{good}} \text{ conductor.}$

Hence in a **good** conductor:

$$\begin{aligned} \tilde{Z}_{\text{cond}}^{\text{good}}(\vec{r}, \omega) &= \mu \left( \frac{\omega}{\sqrt{\mu\omega\sigma_c}} \right) e^{-i(\delta_B(\omega) - \delta_E(\omega))} \hat{z} = \sqrt{\frac{\mu\omega}{\sigma_c}} e^{+i(\delta_E(\omega) - \delta_B(\omega))} \hat{z} = \sqrt{\frac{\varepsilon}{\varepsilon}} \sqrt{\frac{\mu\omega}{\sigma_c}} e^{+i(\delta_E(\omega) - \delta_B(\omega))} \hat{z} \\ &= \sqrt{\frac{\mu}{\varepsilon}} \sqrt{\frac{\varepsilon\omega}{\sigma_c}} e^{+i(\delta_E(\omega) - \delta_B(\omega))} \hat{z} = \left( \sqrt{\frac{\mu}{\varepsilon}} / \sqrt{\frac{\sigma_c}{\varepsilon\omega}} \right) e^{+i(\delta_E(\omega) - \delta_B(\omega))} \hat{z} \quad (\text{Ohms}) \end{aligned}$$

Define {real scalar}:  $Z_{\text{med}}^{\text{lin}} \equiv \sqrt{\frac{\mu}{\varepsilon}}$  Then {real scalar}  $Z_{\text{cond}}^{\text{good}} \equiv \left( \sqrt{\frac{\mu}{\varepsilon}} / \sqrt{\frac{\sigma_c}{\varepsilon\omega}} \right) = Z_{\text{med}}^{\text{lin}} / \left( \frac{\sigma_c}{\varepsilon\omega} \right)^{\frac{1}{2}} \ll Z_{\text{med}}^{\text{lin}}$

since  $\left(\frac{\sigma_c}{\varepsilon\omega}\right) \gg 1$  in a **good** conductor!

Since {in general} the complex impedance is:  $\tilde{Z}(\vec{r}, \omega) = |\tilde{Z}(\vec{r}, \omega)| e^{i\phi_Z(\omega)} \hat{z} = |\tilde{Z}(\vec{r}, \omega)| e^{i(\delta_E(\omega) - \delta_B(\omega))} \hat{z}$

We see the **phase** of the complex wave impedance for a **good** conductor is:

$$\phi_Z(\omega) = \tan^{-1} \left( \frac{-\kappa(\omega)}{k(\omega)} \right) = \tan^{-1}(-1) = -45^\circ = \delta_E(\omega) - \delta_B(\omega) = -(\delta_B(\omega) - \delta_E(\omega)) = -\phi_k(\omega)$$

**Instantaneous EM Wave Energy Densities in a Good Conductor:**

$$\left(\frac{\sigma_C}{\varepsilon\omega}\right) \gg 1$$

$$u_{EM} = u_E^{EM} + u_M^{EM} = \left(\frac{1}{2}\varepsilon E^2\right) + \left(\frac{1}{2\mu}B^2\right) = \left(\frac{1}{2}\varepsilon\vec{E}\cdot\vec{E}\right) + \left(\frac{1}{2\mu}\vec{B}\cdot\vec{B}\right)$$

$$\phi_k \equiv (\delta_B - \delta_E) \approx \frac{\pi}{4} = 45^\circ$$

{in a *good* conductor}

$$\vec{E}(\vec{r}, t) = E_o e^{-\kappa z} \cos(kz - \omega t + \delta_E) \hat{x} \quad \text{and} \quad \vec{B}(\vec{r}, t) = B_o e^{-\kappa z} \cos(kz - \omega t + \delta_E + \phi) \hat{y}$$

Where:  $B_o = \frac{|\tilde{k}(\omega)|}{\omega} E_o = \left[\varepsilon\mu\sqrt{1 + \left(\frac{\sigma_C}{\varepsilon\omega}\right)^2}\right]^{1/2} E_o \approx \sqrt{\frac{\mu\sigma_C}{\varepsilon\omega}} E_o = \sqrt{\frac{\mu\sigma_C}{\omega}} E_o$  for a **good** conductor,

And:  $k(\omega) \approx \kappa(\omega) \approx \sqrt{\frac{\omega\mu\sigma_C}{2}}$ ,

$$v_\phi(\omega) = \frac{\omega}{k(\omega)} \approx \frac{\omega}{\sqrt{\frac{\omega\mu\sigma_C}{2}}} = \sqrt{\frac{2\omega}{\omega\mu\sigma_C}} = \sqrt{\frac{2}{\mu\sigma_C}}$$

for a **good** conductor.

Then:

$$u_E^{EM}(\vec{r}, t) = \frac{1}{2}\varepsilon E^2 = \frac{1}{2}\varepsilon\vec{E}\cdot\vec{E} = \frac{1}{2}\varepsilon E_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E) \quad \text{and:}$$

$$u_M^{EM}(\vec{r}, t) = \frac{1}{2\mu}B^2 = \frac{1}{2\mu}\vec{B}\cdot\vec{B} = \frac{1}{2\mu}B_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E + \phi_k) = \frac{\sigma_C}{2\omega} E_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E + \phi_k)$$

**Time-averaging** these quantities over one complete cycle:

$$\langle u(\vec{r}, t) \rangle \equiv \frac{1}{\tau} \int_0^\tau u(\vec{r}, t) dt$$

$$\langle u_E^{EM}(\vec{r}, t) \rangle = \frac{1}{2}\varepsilon E_o^2 e^{-2\kappa z} \underbrace{\frac{1}{\tau} \int_0^\tau \cos^2(kz - \omega t + \delta_E) d\tau}_{=\frac{1}{2}} = \frac{1}{4}\varepsilon E_o^2 e^{-2\kappa z}$$

$$\langle u_M^{EM}(\vec{r}, t) \rangle = \frac{\sigma_C}{2\omega} E_o^2 e^{-2\kappa z} \underbrace{\frac{1}{\tau} \int_0^\tau \cos^2(kz - \omega t + \delta_E + \phi_k) d\tau}_{=\frac{1}{2}} = \frac{1}{4} \left(\frac{\sigma_C}{\omega}\right) E_o^2 e^{-2\kappa z} = \left(\frac{\sigma_C}{\varepsilon\omega}\right) \cdot \frac{1}{4} \varepsilon E_o^2 e^{-2\kappa z}$$

$$\therefore \langle u_{Tot}^{EM}(\vec{r}, t) \rangle = \langle u_E^{EM}(\vec{r}, t) \rangle + \langle u_M^{EM}(\vec{r}, t) \rangle = \frac{1}{4}\varepsilon \left(1 + \frac{\sigma_C}{\varepsilon\omega}\right) E_o^2 e^{-2\kappa z} \quad \text{n.b. Exponentially attenuated in } z \text{ !!!}$$

But:  $\left(\frac{\sigma_C}{\varepsilon\omega}\right) \gg 1$  for a **good** conductor,  $\Rightarrow \langle u_{Tot}^{EM}(\vec{r}, t) \rangle \approx \frac{1}{2} \left(\frac{\sigma_C}{\varepsilon\omega}\right) \left[\frac{1}{2}\varepsilon E_o^2 e^{-2\kappa z}\right]$

i.e. the **ratio**:  $\frac{\langle u_M^{EM}(\vec{r}, t) \rangle}{\langle u_E^{EM}(\vec{r}, t) \rangle} = \left(\frac{\sigma_C}{\varepsilon\omega}\right) \gg 1$  or:  $\langle u_M^{EM}(\vec{r}, t) \rangle \gg \langle u_E^{EM}(\vec{r}, t) \rangle$  for a **good** conductor.

$\Rightarrow$  Vast majority of EM wave energy is carried by the **magnetic field** in a **good** conductor !!!

Poynting's Vector:  $\vec{S} = \frac{1}{\mu} \vec{E} \times \vec{B} \Rightarrow \langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{\mu} \langle \vec{E} \times \vec{B} \rangle = \frac{1}{2\mu} E_o B_o e^{-2\kappa z} \cos \phi_k \hat{z} \leftarrow \phi_k = \frac{\pi}{4}$

EM wave intensity (aka irradiance):  $I(\vec{r}) = \langle |\vec{S}(\vec{r}, t)| \rangle = \frac{1}{2\mu} E_o B_o e^{-2\kappa z} \cos \phi_k = \frac{1}{2\mu} E_o^2 e^{-2\kappa z} \left( \frac{|\tilde{k}|}{\omega} \cos \phi_k \right)$

But:  $\frac{|\tilde{k}| \cos \phi_k}{\omega} = \frac{k}{\omega} \approx \frac{\sqrt{\frac{\omega \mu \sigma_c}{2}}}{\omega} = \sqrt{\frac{\mu \sigma_c}{2\omega}} \therefore I(\vec{r}) = \langle |\vec{S}(\vec{r}, t)| \rangle = \frac{1}{2\mu} \left( \frac{k}{\omega} \right) E_o^2 e^{-2\kappa z} = \frac{1}{2} \sqrt{\frac{\sigma_c}{2\mu\omega}} E_o^2 e^{-2\kappa z}$

b.) Special/Limiting Case of a Fair Conductor:  $\left( \frac{\sigma_c}{\epsilon\omega} \right) \approx 1 \Rightarrow$  Must use exact formulae!

c.) Special/Limiting Case of a Poor Conductor: (i.e.  $\approx$  an insulator):

Here:  $\left( \frac{\sigma_c}{\epsilon\omega} \right) \ll 1$ . Conductivity of poor conductor:  $\sigma_c \rightarrow 0$  (i.e.  $\rho_c = 1/\sigma_c \rightarrow \infty$ ).

Complex wavenumber:  $\tilde{k} = k + i\kappa$ , with:  $k = k(\omega) = \Re\{\tilde{k}(\omega)\}$  and:  $\kappa = \kappa(\omega) = \Im\{\tilde{k}(\omega)\}$ .

Noting that to 1<sup>st</sup> order in the Taylor series expansion:  $\sqrt{1+\epsilon} \approx 1 + \frac{1}{2}\epsilon$  for  $\epsilon \ll 1$ , thus:

$$k(\omega) \equiv \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} + 1 \right]^{1/2} \approx \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ 1 + \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right)^2 + 1 \right]^{1/2} = \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ 2 + \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right)^2 \right]^{1/2} \approx \omega \sqrt{\epsilon\mu}$$

$$\kappa(\omega) \equiv \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \sqrt{1 + \left( \frac{\sigma_c}{\epsilon\omega} \right)^2} - 1 \right]^{1/2} \approx \omega \sqrt{\frac{\epsilon\mu}{2}} \left[ \cancel{\lambda} + \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right)^2 - \cancel{\lambda} \right]^{1/2} = \omega \sqrt{\frac{\epsilon\mu\sigma_c^2}{4\epsilon^2\omega^2}} \approx \frac{1}{2} \sigma_c \sqrt{\frac{\mu}{\epsilon}}$$

$\therefore k(\omega) \approx \omega \sqrt{\epsilon\mu}$  and:  $\kappa(\omega) \approx \frac{1}{2} \sigma_c \sqrt{\frac{\mu}{\epsilon}}$  for a poor conductor.

In a poor conductor  $\left( \frac{\sigma_c}{\epsilon\omega} \right) \ll 1$ , the ratio:  $\left( \frac{\kappa(\omega)}{k(\omega)} \right) \approx \frac{\frac{1}{2} \sigma_c \sqrt{\frac{\mu}{\epsilon}}}{\omega \sqrt{\epsilon\mu}} = \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right) \ll 1$  i.e.  $\kappa(\omega) \ll k(\omega)$ .

$\Rightarrow$  The complex wavenumber  $\tilde{k} \equiv k + i\kappa$  is primarily real, because  $\kappa \ll k$  in a poor conductor.

Phase angle in a poor conductor:  $\phi_k \equiv \delta_B - \delta_E = \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right) = \tan^{-1} \left( \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right) \right) \approx \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right) \sim 0 \ll 1$

$\Rightarrow \delta_B = \delta_E + \phi_k \approx \delta_E$ , i.e.  $\vec{B}$  and  $\vec{E}$  are nearly in phase with each other in a poor conductor (i.e. dissipation/losses very small in a poor conductor).

In a typical poor conductor, e.g. pure water:

Water has a **huge static** electric permittivity (due to the permanent electric dipole moment of water molecule):  $\epsilon_{H_2O} \approx 81\epsilon_o$  (@  $f = 0 \text{ Hz}$ ) (at  $P = 1 \text{ ATM}$  and  $T = 20^\circ \text{ C}$ ), however, at **optical** frequencies ( $\omega \approx 10^{16} \text{ rad/sec}$ ):  $\epsilon_{H_2O}(\omega) \approx 1.777\epsilon_o$ , where  $\epsilon_o = 8.85 \times 10^{-12} \text{ Farads/m}$ .

Since water is  $\cong$  **non-magnetic**:  $\mu_{H_2O} \approx \mu_o = 4\pi \times 10^{-7} \text{ Henrys/m}$

$\Rightarrow$  index of refraction:  $n_{H_2O}(\omega) = \sqrt{\epsilon_{H_2O}(\omega)\mu_{H_2O}/\epsilon_o\mu_o} \approx 1.333$  at **optical** frequencies.

The conductivity of pure water is:  $\sigma_c^{H_2O} = 1/\rho_c^{H_2O} \approx 1/2.5 \times 10^5 \Omega\text{-m} = 4.0 \times 10^{-6} \text{ Siemens/m}$

(at  $P = 1 \text{ ATM}$  and  $T = 20^\circ \text{ C}$ ). Thus, the criteria for a **poor** conductor ( $\sigma_c/\epsilon\omega$ )  $\approx 2.54 \times 10^{-11} \ll 1$  is certainly satisfied at **optical** frequencies.

The wavenumber in pure  $H_2O$  at **optical** frequencies is:

$$k_{H_2O}(\omega) \approx \omega\sqrt{\epsilon\mu} \approx \omega\sqrt{\epsilon\mu_o} = 10^{16} \sqrt{1.777 \times 8.85 \times 4\pi \times 10^{-7}} \approx 4.45 \times 10^7 \text{ radians/m}$$

The wavelength in pure  $H_2O$  is:  $\lambda_{H_2O} = 2\pi/k_{H_2O} = 1.413 \times 10^{-7} \text{ m} = 141.3 \text{ nm}$  at **optical** frequencies.

cf w/ the **vacuum** wavelength:  $\lambda_o = c/f = 2\pi c/\omega = 1.885 \times 10^{-7} \text{ m} = 188.5 \text{ nm}$

Note that the **optical** wavelength ratio:  $\left(\frac{\lambda_o}{\lambda_{H_2O}}\right) = \frac{188.5 \text{ nm}}{141.3 \text{ nm}} = 1.333 = n_{H_2O}$  for a **poor** conductor.

Skin depth:  $\delta_{sc}(\omega) \equiv \frac{1}{\kappa(\omega)} \approx \frac{1}{\frac{1}{2}\sigma_c\sqrt{\mu/\epsilon}}$  for a **poor** conductor  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$ .

For pure  $H_2O$  at **optical** frequencies:

$$\kappa_{H_2O}(\omega) \approx \frac{1}{2}\sigma_c\sqrt{\frac{\mu}{\epsilon}} \approx \frac{1}{2}\sigma_c\sqrt{\frac{\mu_o}{\epsilon}} = \frac{1}{2}\left(\frac{1}{2.5 \times 10^5}\right)\sqrt{\frac{4\pi \times 10^{-7}}{1.777 \times 8.85 \times 10^{-12}}} \approx 5.65 \times 10^{-4} \text{ rad/m}$$

$$\delta_{sc}^{H_2O}(\omega) \equiv \frac{1}{\kappa_{H_2O}} = 1.7688 \times 10^3 \text{ m} = 1.77 \text{ km}$$

*n.b.* neglects/ignores Rayleigh scattering process – visible light photons elastically scattering off of  $H_2O$  molecules.  $\lambda_{atten}^{vis} \approx 10 \text{ m}$

$$\text{Ratio: } \left(\frac{\kappa_{H_2O}(\omega)}{k_{H_2O}(\omega)}\right) = \frac{\frac{1}{2}\sigma_c\sqrt{\frac{\mu}{\epsilon}}}{\omega\sqrt{\epsilon\mu}} = \frac{1}{2}\left(\frac{\sigma_c}{\epsilon\omega}\right) = \frac{1}{2}\left(\frac{1}{2.5 \times 10^5}\right)\frac{1}{1.777 \times 8.85 \times 10^{-12} \times 10^{16}} = 1.27 \times 10^{-11} \ll 1$$

$$\text{Phase difference: } \phi_k \equiv \delta_B - \delta_E = \tan^{-1}\left(\frac{\kappa_{H_2O}}{k_{H_2O}}\right) \approx 1.27 \times 10^{-11} \text{ radians } (\ll 1) \text{ i.e. } \delta_B = \delta_E + \phi_k \approx \delta_E$$

$\Rightarrow \vec{B}$  and  $\vec{E}$  are **nearly** in phase with each other in pure  $H_2O$  at **optical** frequencies.

For pure  $H_2O$  at low frequencies – e.g. 60 Hz AC line frequency ( $\omega_{AC} = 2\pi f_{AC} = 120\pi \text{ rad/sec}$ ):

The electric permittivity at  $f = 60 \text{ Hz}$  is  $\epsilon_{H_2O}^{AC} (f \approx 60 \text{ Hz}) \approx 80\epsilon_o = 80 \times 8.85 \times 10^{-12} \text{ Farads/m}$   
 and  $\mu_{H_2O}^{AC} \approx \mu_o = 4\pi \times 10^{-7} \text{ Henrys/m}$ . Conductivity of pure  $H_2O$ :  $\sigma_C^{H_2O} = 4.0 \times 10^{-6} \text{ Siemens/m}$

Note that the criteria for a poor conductor:  $\left( \frac{\sigma_C}{\epsilon_{H_2O}^{AC} \omega_{AC}} \right) \approx \frac{4.0 \times 10^{-6}}{80 \times 8.85 \times 10^{-12} \cdot 120\pi} \approx 15 \ll 1$

is not satisfied at the 60 Hz AC line frequency – i.e. at low enough frequencies, even poor conductors such as pure water are actually quite good conductors !!!

Thus, for the following, we must use the good conductor approximations:

$$k_{AC}^{H_2O}(\omega) \approx \kappa_{AC}^{H_2O}(\omega) = \sqrt{\frac{\omega_{AC} \mu_{AC}^{H_2O} \sigma_C}{2}} \approx \sqrt{\frac{\omega_{AC} \mu_o \sigma_C}{2}} = \sqrt{\frac{120\pi \cdot 4\pi \times 10^{-7} \cdot 4 \times 10^{-6}}{2}} = 3.08 \times 10^{-5} \text{ rads/m}$$

$$\lambda_{AC}^{H_2O}(\omega) \approx \frac{2\pi}{k_{AC}^{H_2O}(\omega)} = 2.04 \times 10^5 \text{ m} \quad \text{cf w/ vacuum wavelength: } \lambda_o = c/f_{AC} = \frac{2\pi c}{\omega_{AC}} = 5.00 \times 10^6 \text{ m}$$

Vacuum/good conductor wavelength ratio:  $\left( \frac{\lambda_o}{\lambda_{AC}^{H_2O}} \right) = \frac{5.00 \times 10^6 \text{ m}}{2.04 \times 10^5 \text{ m}} \approx 24.495$

Skin depth for pure  $H_2O$  at 60 Hz AC line frequency:  $\delta_{H_2O}^{AC} \equiv 1/\kappa_{H_2O}^{AC} \approx 3.25 \times 10^4 \text{ m} = 32.5 \text{ km}$

This may seem like a large distance scale associated with the attenuation of the 60 Hz EM waves propagating in pure water, however compare the skin depth to the wavelength at this frequency:  $\delta_{H_2O}^{AC} = 32.5 \text{ km}$  vs.  $\lambda_{AC}^{H_2O} = 1.77 \times 10^6 \text{ m}$ , i.e. we see that  $\delta_{H_2O}^{AC} \ll \lambda_{AC}^{H_2O}$ , as we expect for the case of a good conductor !!!

The ratio  $(\kappa_{H_2O}^{AC}/k_{H_2O}^{AC}) \approx 1$  for pure  $H_2O$  at 60 Hz AC line frequency, which is what we expect for a good conductor {this ratio should be  $\ll 1$  for a poor conductor!}.

Thus, the phase difference is:  $\phi_k \equiv \delta_B - \delta_E = \tan^{-1}(\kappa_{H_2O}^{AC}/k_{H_2O}^{AC}) \approx \tan^{-1}(1) = \frac{\pi}{4} = 45^\circ$

which again is what we expect for a good conductor, i.e.  $\vec{B}$  lags  $\vec{E}$  by  $45^\circ$ !

**Phase speed vs. group speed in a Poor conductor:**

Given that:  $k(\omega) \approx \omega\sqrt{\epsilon\mu}$  and:  $\kappa(\omega) \approx \frac{1}{2}\sigma_c\sqrt{\frac{\mu}{\epsilon}}$  in a **poor** conductor, where:  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$

$\Rightarrow$  The complex wavenumber  $\tilde{k} \equiv k + i\kappa$  is primarily **real**, because  $\kappa \ll k$  in a **poor** conductor.

The **phase** speed in a **poor** conductor is:

$$v_\phi(\omega) \equiv \frac{\omega}{k(\omega)} = \frac{\omega}{\omega\sqrt{\epsilon\mu}} = \frac{1}{\sqrt{\epsilon\mu}} \quad \text{compare to vacuum/free space: } c = \frac{1}{\sqrt{\epsilon_o\mu_o}}$$

The **group** speed in a **poor** conductor is:

$$v_g(\omega) = \left[ \frac{dk(\omega)}{d\omega} \right]^{-1} \approx \frac{1}{\sqrt{\epsilon\mu}} = v_\phi(\omega) !!!$$

**Complex impedance in a Poor conductor:**

$$\tilde{Z}(\vec{r}, \omega) = \mu\omega \left( \frac{\tilde{k}^*(\omega)}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu\omega \left( \frac{|\tilde{k}(\omega)| e^{-i\phi_k(\omega)}}{|\tilde{k}(\omega)|^2} \right) \hat{z} = \mu \left( \frac{\omega}{|\tilde{k}(\omega)|} \right) e^{-i\phi_k(\omega)} \hat{z} \quad (\text{Ohms})$$

But:  $k(\omega) \approx \omega\sqrt{\epsilon\mu}$  and:  $\kappa(\omega) \approx \frac{1}{2}\sqrt{\mu\omega\sigma_c}$  in a **poor** conductor, where:  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$

And:  $\phi_k(\omega) = (\delta_B - \delta_E) = \tan^{-1} \left( \frac{\kappa(\omega)}{k(\omega)} \right) = \tan^{-1} \left( \frac{1}{2} \left( \frac{\omega_c}{\omega} \right) \right) \approx \frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right) \sim 0 \ll 1$  E and B are  $\underline{\text{in-phase}}$  with each other for a **poor** conductor.

Hence in a **poor** conductor:  $\tilde{Z}_{cond}^{poor}(\vec{r}, \omega) \approx \sqrt{\frac{\mu}{\epsilon}} \hat{z} = Z_{cond}^{poor} \hat{z}$ . Define {real scalar}:  $Z_{med}^{lin} \equiv \sqrt{\frac{\mu}{\epsilon}}$ .

Then {real scalar} characteristic longitudinal **EM** wave impedance for a **poor** conductor is:

$$Z_{cond}^{poor} = Z_{med}^{lin} = \sqrt{\frac{\mu}{\epsilon}} \quad (\text{Ohms}) \quad \text{n.b. Compare to free space: } Z_o \equiv \sqrt{\frac{\mu_o}{\epsilon_o}} \approx 376.8 \quad (\text{Ohms})$$

Since {in general} the complex impedance is:  $\tilde{Z}(\vec{r}, \omega) = |\tilde{Z}(\vec{r}, \omega)| e^{i\phi_Z(\omega)} \hat{z} = |\tilde{Z}(\vec{r}, \omega)| e^{i(\delta_E(\omega) - \delta_B(\omega))} \hat{z}$ .

The **phase** of the complex wave impedance for a **poor** conductor is:

$$\phi_Z(\omega) = \tan^{-1} \left( \frac{-\kappa(\omega)}{k(\omega)} \right) = \tan^{-1} \left( -\frac{1}{2} \left( \frac{\sigma_c}{\epsilon\omega} \right) \right) \approx 0^\circ = \delta_E(\omega) - \delta_B(\omega)$$

**Instantaneous EM energy densities in a poor conductor:**  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$

$$u_{EM}(\vec{r}, t) = u_E^{EM}(\vec{r}, t) + u_M^{EM}(\vec{r}, t) = \left(\frac{1}{2}\epsilon E^2\right) + \left(\frac{1}{2\mu}B^2\right) = \left(\frac{1}{2}\epsilon\vec{E}\cdot\vec{E}\right) + \left(\frac{1}{2\mu}\vec{B}\cdot\vec{B}\right)$$

The physical/instantaneous purely real **time-domain**  $\vec{E}$  and  $\vec{B}$  fields are:

$$\vec{E}(\vec{r}, t) = E_o e^{-\kappa z} \cos(kz - \omega t + \delta_E) \hat{x} \quad \text{and:} \quad \vec{B}(\vec{r}, t) = B_o e^{-\kappa z} \cos(kz - \omega t + \delta_E + \phi_k) \hat{y}$$

where:  $B_o = \frac{|\tilde{k}|}{\omega} E_o = \left[\epsilon\mu\sqrt{1 + \left(\frac{\sigma_c}{\epsilon\omega}\right)^2}\right]^{1/2} E_o \approx \sqrt{\epsilon\mu} E_o$  for a **poor** conductor,  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$ .

$k \approx \omega\sqrt{\epsilon\mu} = \frac{\omega}{v_\phi}$  where:  $v_\phi = \frac{\omega}{k(\omega)} = \frac{1}{\sqrt{\epsilon\mu}}$  for a **poor** conductor.

and:  $\kappa \approx \frac{1}{2}\sigma_c\sqrt{\frac{\mu}{\epsilon}} \ll k \approx \omega\sqrt{\epsilon\mu}$ ,  $|\tilde{k}| \approx \omega\sqrt{\epsilon\mu}$  for a **poor** conductor.

then:  $u_E^{EM}(\vec{r}, t) = \frac{1}{2}\epsilon E^2 = \frac{1}{2}c\vec{E}\cdot\vec{E} = \frac{1}{2}\epsilon E_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E)$  and:

$$u_M^{EM}(\vec{r}, t) = \frac{1}{2\mu}B^2 = \frac{1}{2\mu}\vec{B}\cdot\vec{B} = \frac{1}{2\mu}B_o^2 e^{-2\kappa z} \cos^2(kz - \omega t + \delta_E + \phi_k)$$

**Time-averaging** these quantities:

$$\langle u_E^{EM}(\vec{r}, t) \rangle = \frac{1}{4}\epsilon E_o^2 e^{-2\kappa z} \quad \text{and:} \quad \langle u_M^{EM}(\vec{r}, t) \rangle = \frac{1}{4\mu}B_o^2 e^{-2\kappa z} \approx \frac{1}{4\cancel{\mu}}(\epsilon\cancel{\mu})E_o^2 e^{-2\kappa z} = \frac{1}{4}\epsilon E_o^2 e^{-2\kappa z}$$

$$\therefore \langle u_{Tot}^{EM}(\vec{r}, t) \rangle = \langle u_E^{EM}(\vec{r}, t) \rangle + \langle u_M^{EM}(\vec{r}, t) \rangle \approx \frac{1}{4}\epsilon E_o^2 e^{-2\kappa z} + \frac{1}{4}\epsilon E_o^2 e^{-2\kappa z} = \frac{1}{2}\epsilon E_o^2 e^{-2\kappa z}$$

Thus:  $\langle u_{Tot}^{EM}(\vec{r}, t) \rangle = \frac{1}{2}\epsilon E_o^2 e^{-2\kappa z}$  for a **poor** conductor,  $\left(\frac{\sigma_c}{\epsilon\omega}\right) \ll 1$ .

The **ratio** of {time-averaged} electric/magnetic energy densities for a **poor** conductor:

$$\frac{\langle u_E^{EM}(\vec{r}, t) \rangle}{\langle u_M^{EM}(\vec{r}, t) \rangle} \approx \frac{\frac{1}{4}\epsilon E_o^2 e^{-2\kappa z}}{\frac{1}{4}\epsilon E_o^2 e^{-2\kappa z}} = 1 \quad \phi_k \equiv \delta_B - \delta_E = \tan^{-1}\left(\frac{\kappa_{H_2O}}{k_{H_2O}}\right) \ll 1 \quad \kappa_{H_2O} \approx \frac{1}{2}\sigma_c\sqrt{\frac{\mu_o}{\epsilon}} \ll k_{H_2O} \approx \omega\sqrt{\epsilon\mu_o}$$

$\Rightarrow$  EM wave energy is shared  $\approx$  equally by the  $\vec{E}$  and  $\vec{B}$  fields in a **poor** conductor!

**Instantaneous Poynting's Vector** for EM waves propagating in a **poor** conductor:

$$\vec{S}(\vec{r}, t) = \frac{1}{\mu}\vec{E}(\vec{r}, t) \times \vec{B}(\vec{r}, t) \Rightarrow \langle \vec{S}(\vec{r}, t) \rangle = \frac{1}{\mu}\langle \vec{E}(\vec{r}, t) \times \vec{B}(\vec{r}, t) \rangle \approx \frac{1}{2}\sqrt{\frac{\epsilon}{\mu_o}}E_o^2 e^{-2\kappa z} \underbrace{\cos\phi_k}_{\approx 1} \hat{z}$$

**Intensity** of EM waves propagating in a **poor** conductor: 
$$I(\vec{r}) = \langle |\vec{S}(\vec{r}, t)| \rangle = \frac{1}{2} \sqrt{\frac{\epsilon}{\mu_0}} E_o^2 e^{-2\kappa z}$$

### Reflection of EM Waves at Normal Incidence from a Conducting Surface:

In the presence of free surface charges  $\sigma_{free}$  and/or free surface currents,  $\vec{K}_{free}$  the boundary conditions obtained from (the integral forms of) Maxwell's equations for reflection and refraction at e.g. a dielectric-conductor interface become:

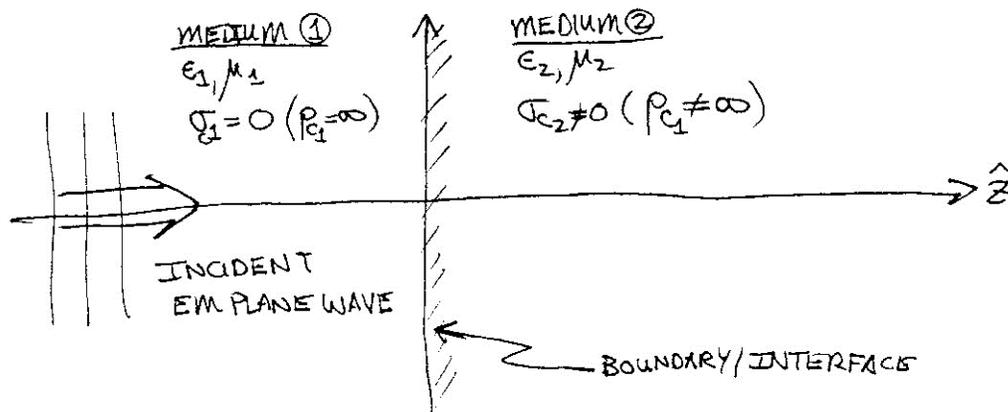
BC 1): (normal $\vec{D}$ at interface):	$\epsilon_1 E_1^\perp - \epsilon_2 E_2^\perp = \sigma_{free}$	$\perp$ = normal to plane of interface $\parallel$ = parallel to plane of interface
BC 2): (tangential $\vec{E}$ at interface):	$E_1^\parallel - E_2^\parallel = 0 \Rightarrow E_1^\parallel = E_2^\parallel$	
BC 3): (normal $\vec{B}$ at interface):	$B_1^\perp - B_2^\perp = 0 \Rightarrow B_1^\perp = B_2^\perp$	
BC 4): (tangential $\vec{H}$ at interface):	$\frac{1}{\mu_1} B_1^\parallel - \frac{1}{\mu_2} B_2^\parallel = \vec{K}_{free} \times \hat{n}_{21}$	

where  $\hat{n}_{21}$  is a unit vector  $\perp$  to the interface, pointing **from** medium (2) **into** medium (1).

{n.b. do **not** confuse  $\hat{n}_{21}$  with the EM wave **polarization vector**  $\hat{n}$  !!!}

Note: For **Ohmic** conductors (i.e. "normal" conductors obeying Ohm's law  $\vec{J}_{free} = \sigma_c \vec{E}$ ) there can be **no free surface** currents - i.e.  $\vec{K}_{free} = 0$ , because  $\vec{K}_{free} \neq 0$  would require an **infinite**  $\vec{E}$ -field at the boundary/interface! ( $\vec{J}_{free} \neq 0$  **inside** the conductor is fine/OK...)

Suppose  $\exists$  a boundary/interface (located in the  $x$ - $y$  plane at  $z = 0$ ) between a non-conducting linear/homogeneous/isotropic medium (1) and a conductor (2). A monochromatic plane EM wave is incident on the interface, linearly polarized in  $+\hat{x}$  direction, traveling in the  $+\hat{z}$  direction, approaches the interface/boundary from the left {in medium (1)} as shown in the figure below:



$$\tilde{\tilde{\mathbf{B}}} = \frac{1}{v} (\hat{\mathbf{k}} \times \tilde{\tilde{\mathbf{E}}})$$

Incident  $EM$  wave {medium (1)}:  $\tilde{\tilde{\mathbf{E}}}_{inc}(\vec{r}, t) = \tilde{\tilde{E}}_{o_{inc}} e^{i(k_1 z - \omega t)} \hat{\mathbf{x}}$  and:  $\tilde{\tilde{\mathbf{B}}}_{inc}(\vec{r}, t) = \frac{1}{v_1} \tilde{\tilde{E}}_{o_{inc}} e^{i(k_1 z - \omega t)} \hat{\mathbf{y}}$

Reflected  $EM$  wave {medium (1)}:  $\tilde{\tilde{\mathbf{E}}}_{refl}(\vec{r}, t) = \tilde{\tilde{E}}_{o_{refl}} e^{i(-k_1 z - \omega t)} \hat{\mathbf{x}}$  and:  $\tilde{\tilde{\mathbf{B}}}_{refl}(\vec{r}, t) = -\frac{1}{v_1} \tilde{\tilde{E}}_{o_{refl}} e^{i(-k_1 z - \omega t)} \hat{\mathbf{y}}$

Transmitted  $EM$  wave {medium (2)}:  $\tilde{\tilde{\mathbf{E}}}_{trans}(\vec{r}, t) = \tilde{\tilde{E}}_{o_{trans}} e^{i(\tilde{k}_2 z - \omega t)} \hat{\mathbf{x}}$  and:  $\tilde{\tilde{\mathbf{B}}}_{trans}(\vec{r}, t) = \frac{\tilde{k}_2}{\omega} \tilde{\tilde{E}}_{o_{trans}} e^{i(\tilde{k}_2 z - \omega t)} \hat{\mathbf{y}}$

*n.b.* **complex** wavenumber in {conducting} medium (2):  $\tilde{k}_2 = k_2 + i\kappa_2$

In medium (1)  $EM$  fields are:  $\tilde{\tilde{\mathbf{E}}}_{Tot_1}(\vec{r}, t) = \tilde{\tilde{\mathbf{E}}}_{inc}(\vec{r}, t) + \tilde{\tilde{\mathbf{E}}}_{refl}(\vec{r}, t)$  and:  $\tilde{\tilde{\mathbf{B}}}_{Tot_1}(\vec{r}, t) = \tilde{\tilde{\mathbf{B}}}_{inc}(\vec{r}, t) + \tilde{\tilde{\mathbf{B}}}_{refl}(\vec{r}, t)$

In medium (2)  $EM$  fields are:  $\tilde{\tilde{\mathbf{E}}}_{Tot_2}(\vec{r}, t) = \tilde{\tilde{\mathbf{E}}}_{trans}(\vec{r}, t)$  and:  $\tilde{\tilde{\mathbf{B}}}_{Tot_2}(\vec{r}, t) = \tilde{\tilde{\mathbf{B}}}_{trans}(\vec{r}, t)$

Apply BC's at the  $z = 0$  interface in the  $x$ - $y$  plane:

BC 1):  $\epsilon_1 E_1^\perp - \epsilon_2 E_2^\perp = \sigma_{free}$  but:  $E_1^\perp = \tilde{E}_{1z} = 0$  and:  $E_2^\perp = \tilde{E}_{2z} = 0 \therefore 0 - 0 = \sigma_{free} \Rightarrow \sigma_{free} = 0$

BC 2):  $E_1^\parallel = E_2^\parallel \therefore \tilde{\tilde{E}}_{o_{inc}} + \tilde{\tilde{E}}_{o_{refl}} = \tilde{\tilde{E}}_{o_{trans}}$

BC 3):  $B_1^\perp = B_2^\perp$  but:  $B_1^\perp = B_{1z} = 0$  and:  $B_2^\perp = B_{2z} = 0 \Rightarrow 0 = 0$

BC 4):  $\frac{1}{\mu_1} B_1^\parallel - \frac{1}{\mu_2} B_2^\parallel = \vec{K}_{free} \times \hat{n}_{21}$  but:  $\vec{K}_{free} = 0 \therefore \frac{1}{\mu_1 v_1} (\tilde{\tilde{E}}_{o_{inc}} - \tilde{\tilde{E}}_{o_{refl}}) - \frac{\tilde{k}_2}{\mu_2 \omega} \tilde{\tilde{E}}_{o_{trans}} = 0$

or:  $\tilde{\tilde{E}}_{o_{inc}} - \tilde{\tilde{E}}_{o_{refl}} = \tilde{\beta} \tilde{\tilde{E}}_{o_{trans}}$  with:  $\tilde{\beta} \equiv \left( \frac{\mu_1 v_1 \tilde{k}_2}{\mu_2 \omega} \right) = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2$

Thus we obtain:  $\tilde{\tilde{E}}_{o_{refl}} = \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right) \tilde{\tilde{E}}_{o_{inc}}$  and:  $\tilde{\tilde{E}}_{o_{trans}} = \frac{2}{(1 + \tilde{\beta})} \tilde{\tilde{E}}_{o_{inc}}$

or:  $\left( \frac{\tilde{\tilde{E}}_{o_{refl}}}{\tilde{\tilde{E}}_{o_{inc}}} \right) = \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right)$  and:  $\left( \frac{\tilde{\tilde{E}}_{o_{trans}}}{\tilde{\tilde{E}}_{o_{inc}}} \right) = \frac{2}{(1 + \tilde{\beta})}$  with:  $\tilde{\beta} \equiv \left( \frac{\mu_1 v_1 \tilde{k}_2}{\mu_2 \omega} \right) = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2$

Note that these ‘‘Fresnel’’ relations for reflection/transmission of  $EM$  waves at normal incidence on a non-conductor/conductor boundary/interface are **identical** to those obtained for reflection / transmission of  $EM$  waves at normal incidence on a boundary/interface between two **linear** non-conductors, **except** for the replacement of  $\beta$  with a now complex  $\tilde{\beta}$  for the present situation.

Note also that **here**,  $\tilde{\beta}$  is **frequency-dependent**, i.e.  $\tilde{\beta} = \tilde{\beta}(\omega) \equiv \left( \frac{\mu_1 v_1 \tilde{k}_2}{\mu_2 \omega} \right) = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2(\omega)$ .

For the case of a **perfect** conductor, the conductivity  $\sigma_C = \infty$  {thus resistivity,  $\rho_C = 1/\sigma_C = 0$ }

$$\Rightarrow \text{both } k_2 \approx \kappa_2 \approx \sqrt{\frac{\mu_2 \omega \sigma_C}{2}} = \infty \text{ and since: } \tilde{k}_2 = k_2 + i\kappa_2 \text{ then: } \tilde{k}_2 = \infty + i\infty = \infty(1+i)$$

$$\text{and since: } \tilde{\beta} \equiv \left( \frac{\mu_1 v_1 \tilde{k}_2}{\mu_2 \omega} \right) = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2 \Rightarrow \tilde{\beta} = \infty$$

$$\text{Thus, for a } \textbf{perfect} \text{ conductor, we see that: } \tilde{E}_{o_{refl}} = -\tilde{E}_{o_{inc}} \text{ and: } \tilde{E}_{trans} = 0$$

Thus, for a **perfect** conductor, the reflection and transmission coefficients are:

$$R \equiv \left( \frac{E_{o_{refl}}}{E_{o_{inc}}} \right)^2 = \left| \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right|^2 = \left( \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right) \left( \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right)^* = 1 \quad \text{and: } T = 1 - R = 0$$

We also see that for a **perfect** conductor, for normal incidence, the **reflected** wave undergoes a  $180^\circ$  **phase shift** with respect to the **incident** wave at the interface/boundary at  $z = 0$  in the  $x$ - $y$  plane. A **perfect** conductor screens out **all** **EM** waves from propagating in its interior.

For the case of a **good** conductor, the conductivity  $\sigma_C$  is finite-large, but not infinite.

The reflection coefficient  $R$  for monochromatic plane **EM** waves at normal incidence on a **good** conductor is **not** unity, but **very** close to it. {This is why **good** conductors make **good** mirrors!}

$$\text{For a } \textbf{good} \text{ conductor: } R \equiv \left( \frac{E_{o_{refl}}}{E_{o_{inc}}} \right)^2 = \left| \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right|^2 = \left( \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right) \left( \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right)^* = \left| \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right|^2 = \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right) \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right)^*$$

$$\text{Where: } \tilde{\beta} \equiv \left( \frac{\mu_1 v_1 \tilde{k}_2}{\mu_2 \omega} \right) = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2 \quad \text{and: } \tilde{k}_2 = k_2 + i\kappa_2. \quad \text{For a } \textbf{good} \text{ conductor: } k_2 \approx \kappa_2 \approx \sqrt{\frac{\mu_2 \omega \sigma_C}{2}}$$

$$\text{Thus: } \tilde{\beta} = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \tilde{k}_2 = \left( \frac{\mu_1 v_1}{\mu_2 \omega} \right) \sqrt{\frac{\mu_2 \omega \sigma_C}{2}} (1+i) = \mu_1 v_1 \sqrt{\frac{\sigma_C}{2\mu_2 \omega}} (1+i)$$

$$\text{Define: } \gamma \equiv \mu_1 v_1 \sqrt{\frac{\sigma_C}{2\mu_2 \omega}} \quad \text{Then: } \tilde{\beta} = \gamma(1+i)$$

Thus, the reflection coefficient  $R$  for monochromatic plane **EM** waves at normal incidence on a **good** conductor is {*n.b.* **frequency-dependent!**}:

$$R = \left| \frac{\tilde{E}_{o_{refl}}}{\tilde{E}_{o_{inc}}} \right|^2 = \left| \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right|^2 = \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right) \left( \frac{1 - \tilde{\beta}}{1 + \tilde{\beta}} \right)^* = \left( \frac{1 - \gamma - i\gamma}{1 + \gamma + i\gamma} \right) \left( \frac{1 - \gamma + i\gamma}{1 + \gamma - i\gamma} \right) = \left[ \frac{(1 - \gamma)^2 + \gamma^2}{(1 + \gamma)^2 + \gamma^2} \right]$$

$$\text{with: } \gamma \equiv \mu_1 v_1 \sqrt{\frac{\sigma_C}{2\mu_2 \omega}} = \gamma(\omega)$$

Obviously, only a {very} **small** fraction of the normally-incident monochromatic plane *EM* wave is **transmitted** into the **good** conductor, since  $R \lesssim 1$  and  $T = 1 - R$ , *i.e.*:

$$T = 1 - R = 1 - \left[ \frac{(1 - \gamma)^2 + \gamma^2}{(1 + \gamma)^2 + \gamma^2} \right] (\ll 1) \quad \text{with:} \quad \gamma \equiv \mu_1 v_1 \sqrt{\frac{\sigma_c}{2\mu_2 \omega}} = \gamma(\omega)$$

Note that the **transmitted** wave is **exponentially** attenuated in the  $z$ -direction; the  $\vec{E}$  and  $\vec{B}$  fields in the **good** conductor fall to  $1/e$  of their initial  $\{z = 0\}$  values (at/on the interface) after the monochromatic plane *EM* wave propagates a distance of one skin depth in  $z$  into the conductor:

$$\delta_{sc}(\omega) \equiv \frac{1}{\kappa_2(\omega)} \approx \sqrt{\frac{2}{\mu_2 \omega \sigma_c}}$$

Note also that the **energy** associated with the **transmitted** monochromatic plane *EM* wave is ultimately dissipated in the conducting medium as **heat**.

In {bulk} metals, since the transmitted wave is {rapidly} absorbed/attenuated in the metal, experimentally we are only able to study/measure the reflection coefficient  $R$ . A full/detailed mathematical description of the physics of reflection from the surface of a metal conductor as a function of angle of incidence *i.e.*  $R(\omega, \theta_{inc})$  and also requires the use of a complex dispersion relation  $\tilde{k}(\omega) = \omega/\tilde{v}(\omega) = (\omega/c)\tilde{n}(\omega)$  with complex  $\tilde{k}(\omega) = k(\omega) + i\kappa(\omega)$  **and** a complex propagation speed  $\tilde{v}(\omega) = v(\omega) + iv(\omega) = c/\tilde{n}(\omega)$  with accompanying complex index of refraction  $\tilde{n}(\omega) = n(\omega) + i\eta(\omega)$ , and is hence commensurately more mathematically complicated....

So-called **ellipsometry** measurements of the *EM* radiation reflected from the surface of the metal as a function of angle of incidence yields information on the real and imaginary parts of the complex index of refraction of the metal  $\tilde{n}(\omega) = n(\omega) + i\eta(\omega)$ , and thus the real and imaginary parts of the complex dielectric constant and/or the complex electric susceptibility of the metal, since  $\tilde{n}(\omega) = \sqrt{\tilde{\epsilon}(\omega)/\epsilon_0} = \sqrt{1 + \tilde{\chi}_e(\omega)}$  or  $\tilde{n}^2(\omega) = \tilde{\epsilon}(\omega)/\epsilon_0 = 1 + \tilde{\chi}_e(\omega)$ .

If interested in learning more about this, *e.g.* please see/read Physics 436 Lect. Notes 8, and *e.g.* please see/read **Optics**, M.V. Klein, p. 588-592, Wiley, 1970 {P436 reference book on reserve in the Physics library}. Please also see/read the UIUC P402 Optics/Light Lab Ellipsometry Lab Handout C4 and **especially** the references at the end. Available at: <http://online.physics.uiuc.edu/courses/phys402/exp/C4/C4.pdf>

We will discuss the **dispersive** nature of dielectric, **non**-conducting materials in the next lecture...

But first, we need to remind the reader of the full Maxwell's equations in matter...

## Full Maxwell Equations in Matter:

The electromagnetic state of matter at a given observation point  $\vec{r}$  at a given time  $t$  is described by four **macroscopic** quantities:

- 1.) The volume density of free charge:  $\rho_{free}(\vec{r}, t) \leftarrow aka \{free\} \text{ charge density}$
- 2.) The volume density of electric dipole moments:  $\vec{P}(\vec{r}, t) \leftarrow aka \text{ electric polarization}$
- 3.) The volume density of magnetic dipole moments:  $\vec{M}(\vec{r}, t) \leftarrow aka \text{ magnetization}$
- 4.) The free electric current/unit area:  $\vec{J}_{free}(\vec{r}, t) \leftarrow aka \{free\} \text{ current density}$

All four of these quantities are macroscopically averaged - *i.e.* the microscopic fluctuations due to atomic/molecular makeup of matter have been smoothed out.

The four above quantities are related to the macroscopic  $\vec{E}$  and  $\vec{B}$  fields by the four Maxwell equations for matter (see Physics 435 Lect. Notes 24, p. 14):

- 1) Gauss' Law: 
$$\vec{\nabla} \cdot \vec{E} = \frac{\rho_{Tot}}{\epsilon_o} = \frac{1}{\epsilon_o} (\rho_{free} + \rho_{bound}), \text{ where: } \rho_{bound} = -\vec{\nabla} \cdot \vec{P}$$

Auxiliary relation:  $\vec{D} = \epsilon_o \vec{E} + \vec{P}$  & constitutive relation:  $\vec{D} = \epsilon \vec{E}$

Electric polarization  $\vec{P} = (\epsilon - \epsilon_o) \vec{E} = \epsilon_o \chi_e \vec{E}$ , electric susceptibility:  $\chi_e = \left( \frac{\epsilon}{\epsilon_o} - 1 \right)$

$$\vec{\nabla} \cdot \vec{D} = \epsilon_o \vec{\nabla} \cdot \vec{E} + \vec{\nabla} \cdot \vec{P} = \rho_{free}$$
- 2) No magnetic charges/monopoles:  $\vec{\nabla} \cdot \vec{B} = 0$ 

Auxiliary relation:  $\vec{H} = \frac{1}{\mu_o} \vec{B} - \vec{M} \Rightarrow \vec{\nabla} \cdot \vec{H} = -\vec{\nabla} \cdot \vec{M}$  & constitutive relation:  $\vec{B} = \mu \vec{H}$
- 3) Faraday's Law: 
$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} = -\mu_o \frac{\partial \vec{H}}{\partial t} - \mu_o \frac{\partial \vec{M}}{\partial t}$$

Magnetization:  $\vec{M} = \left( \frac{\mu}{\mu_o} - 1 \right) \vec{H} = \chi_m \vec{H}$ , magnetic susceptibility:  $\chi_m = \left( \frac{\mu}{\mu_o} - 1 \right)$
- 4) Ampere's Law: 
$$\vec{\nabla} \times \vec{B} = \mu_o \vec{J}_{Tot} + \mu_o \vec{J}_D \text{ with: } \vec{J}_D = \epsilon_o \frac{\partial \vec{E}}{\partial t}$$

Total current density:  $\vec{J}_{Tot} = \vec{J}_{free} + \vec{J}_{bound}^{mag} + \vec{J}_{bound}^P$   $\vec{J}_{bound}^{mag} = \vec{\nabla} \times \vec{M}$   $\vec{J}_{bound}^P = \frac{\partial \vec{P}}{\partial t}$

$$\vec{\nabla} \times \vec{B} = \mu_o \vec{J}_{free} + \mu_o \vec{\nabla} \times \vec{M} + \mu_o \frac{\partial \vec{P}}{\partial t} + \mu_o \epsilon_o \frac{\partial \vec{E}}{\partial t}$$

$$\vec{\nabla} \times \vec{H} = \mu_o \vec{J}_{free} + \mu_o \frac{\partial \vec{D}}{\partial t}$$

We also have Ohm's Law:  $\vec{J} = \sigma_c \vec{E}$  and the 3 continuity equation(s):  $\vec{\nabla} \cdot \vec{J}_\alpha = -\frac{\partial \rho_\alpha}{\partial t}$  associated with subscript index  $\alpha$  for  $\alpha = \text{free, bound and total}$  electric charge conservation.

For many/most (but not all!!!) physics problems, *e.g.* in optics/condensed matter physics, materials of interest are frequently **non-magnetic** (or **negligibly** magnetic) and have **no** free charge densities present, *i.e.*  $\rho_{free} = 0$ . If  $\mu \approx \mu_o$ , then:  $\vec{M} = 0$  and thus:  $\vec{H} = \vec{B}/\mu_o$  in such non-magnetic materials.

Then Maxwell's equations in matter, for  $\rho_{free} = 0$  and  $\vec{M} = 0$  reduce to:

1) Gauss' Law:  $\vec{\nabla} \cdot \vec{D} = 0$  or:  $\vec{\nabla} \cdot \vec{E} = -\frac{1}{\epsilon_o} \vec{\nabla} \cdot \vec{P} = \frac{\rho_{bound}}{\epsilon_o}$

2) No magnetic charges:  $\vec{\nabla} \cdot \vec{B} = 0$

3) Faraday's Law:  $\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$

4) Ampere's Law:  $\vec{\nabla} \times \vec{B} = \mu_o \epsilon_o \frac{\partial \vec{E}}{\partial t} + \mu_o \frac{\partial \vec{P}}{\partial t} + \mu_o \vec{J}_{free}$

We also have Ohm's Law  $\vec{J}_{free} = \sigma_c \vec{E}$  and the Continuity eqn.  $\vec{\nabla} \cdot \vec{J}_{free} = 0$  {here}.

Then applying the curl operator to Faraday's Law:

$$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = -\frac{\partial}{\partial t} (\vec{\nabla} \times \vec{B}) = -\mu_o \epsilon_o \frac{\partial^2 \vec{E}}{\partial t^2} - \mu_o \frac{\partial^2 \vec{P}}{\partial t^2} - \mu_o \frac{\partial \vec{J}_{free}}{\partial t} = \vec{\nabla} (\vec{\nabla} \cdot \vec{E}) - \nabla^2 \vec{E} = \frac{1}{\epsilon_o} \vec{\nabla} \rho_{bound} - \nabla^2 \vec{E}$$

We obtain the **inhomogeneous** wave equation:

$$\nabla^2 \vec{E} - \frac{1}{c^2} \frac{\partial^2 \vec{E}}{\partial t^2} = \underbrace{\frac{1}{\epsilon_o} \vec{\nabla} \rho_{bound} + \mu_o \frac{\partial^2 \vec{P}}{\partial t^2} + \mu_o \frac{\partial \vec{J}_{free}}{\partial t}}_{\text{source terms}} \quad \{\text{and a similar/analogous one for } \vec{B}\}$$

For nonconducting/poorly-conducting media, *i.e.* insulators/dielectrics, the first two terms on the RHS of the above equation are important – *e.g.* they explain many optical effects such as dispersion (wavelength/frequency-dependence of the index of refraction), absorption, double – refraction/bi-refringence, optical activity, . . . .

Note that the  $\vec{\nabla} \rho_{bound} = -\vec{\nabla} (\vec{\nabla} \cdot \vec{P})$  term is often zero, *e.g.* if the electric polarization  $\vec{P}$  is uniform:

where:  $\vec{\nabla} = \frac{\partial}{\partial x} \hat{x} + \frac{\partial}{\partial y} \hat{y} + \frac{\partial}{\partial z} \hat{z}$  and:  $\vec{\nabla} \cdot \vec{P} = \frac{\partial P_x}{\partial x} + \frac{\partial P_y}{\partial y} + \frac{\partial P_z}{\partial z} = 0$  if  $\vec{P}$  is uniform

or, if the polarization  $\vec{P} \propto \vec{E}$ , where *e.g.*  $\vec{E}(\vec{r}, t) = E_o \cos(kz - \omega t + \delta) \hat{x}$ , then:  $\vec{\nabla} \cdot \vec{P} = 0$  also.

For **good** conductors (*e.g.* metals), the **conduction term**  $\mu_o \frac{\partial \vec{J}_{free}}{\partial t} = \mu_o \sigma_c \frac{\partial \vec{E}}{\partial t}$  is the most important, because it explains the **opacity** of metals (*e.g.* in the visible light region) and also explains the **high reflectance** of metals.

**All** source terms on the RHS of the above inhomogeneous wave equation are of importance for **semi-conductors** – however a **proper/more complete** physics description of *EM* wave propagation in semiconductors also requires the addition of quantum theory for rigorous treatment...